



Cell-free massive MIMO Networks

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Important concepts of wireless channels







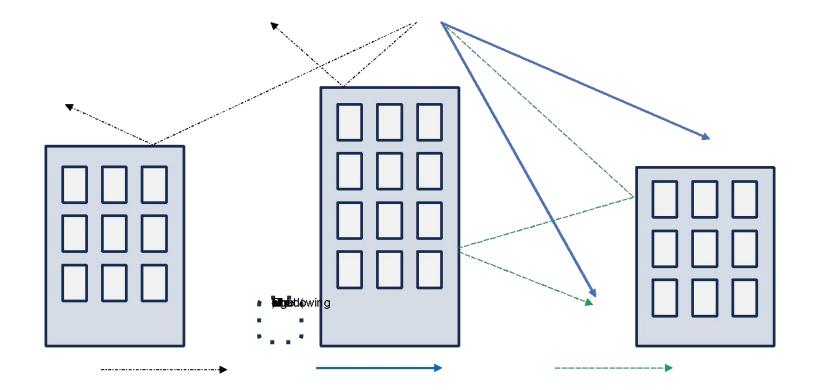
Channel propagation



Line-of-sight (LoS) propagation

Non-Line-of-Sight (NLoS) propagation (multipath)

Shadowing



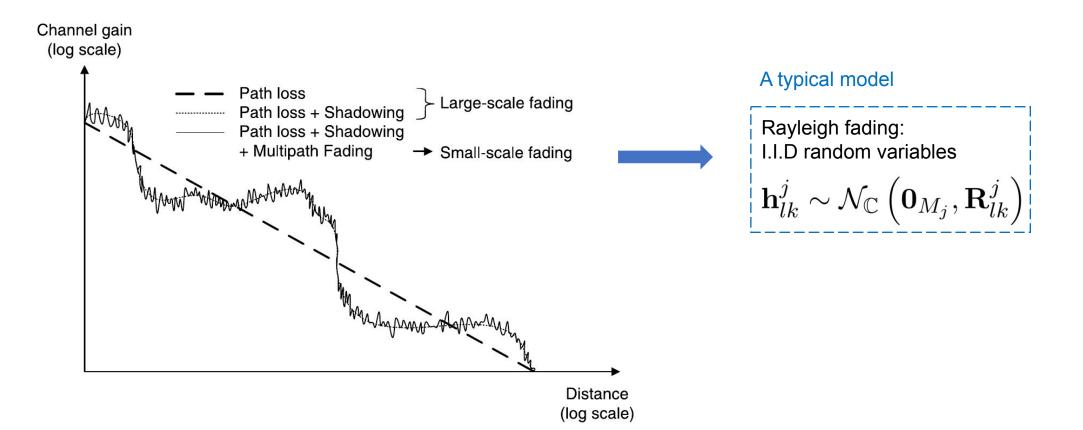


Channel fading



Fading

- Small-Scale Fading: rapid fluctuations in signal strength over short distances and short periods of time
- Large-Scale Fading: variations in signal strength over longer distances and time periods



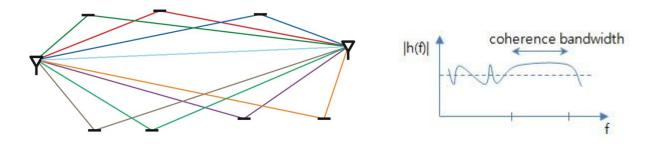


Coherence Time and Bandwidth



Coherence Block

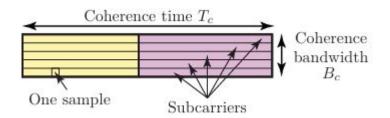
Coherence bandwidth
 Multipath Delays → phase shifts in frequency → channel changes in frequency
 (→ coherence bandwidth)

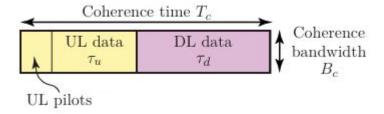


Coherence time
 Doppler shift → phase shift in time → channel changes in time (→ coherence time)



Coherence block









Cellular networks, MIMO and co-located massive MIMO



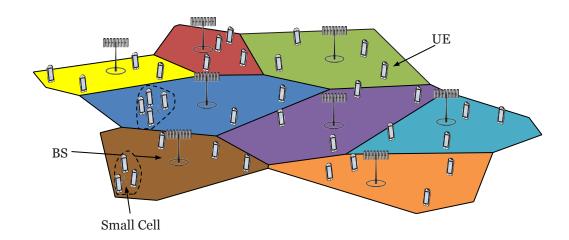


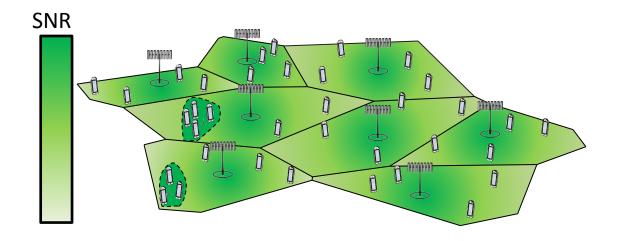


Cellular Networks



- The coverage area is called cell
- A base station (BS) serves a set of user equipment (UE)





Main tiers

- Coverage tier: outdoor
- Hotspot tier: indoor and limited area outdoor service



Cellular Networks

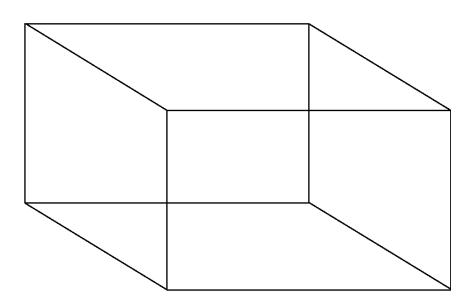


The concept of area throughput

- Modeled as the volume of a rectangular box
- Measured in bit/s/Km²

A question arises

- What should we improve? CD, BW or SE?
- What are the impacts of improving CD, BW or SE?





Cellular Networks



Increasing BW

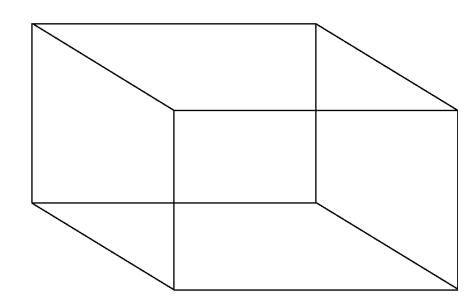
- Impaired by license and spectrum sharing
- Very hard to do for the coverage tier

Increasing CD

There are some challenges...

Increasing SE

Allows to use the BSs and bandwidth more effectively





Ways to improve the Spectral Efficiency (SE)



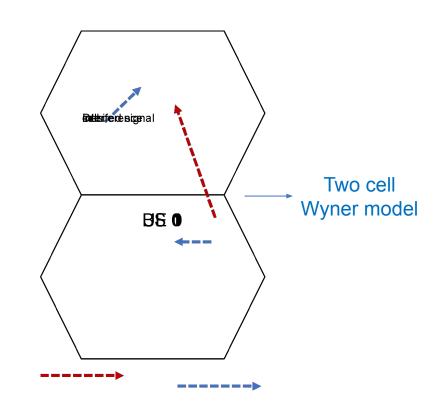
Increasing the transmit power

- Pushes the network into an interference-limited regime
- Example, a Line-of-Sight (LOS) case. Two cells and two UEs

$$\mathsf{SE}_0^{\mathrm{LoS}} = \log_2\left(1 + \frac{1}{\bar{\beta} + \frac{1}{\mathsf{SNR}_0}}\right),\$$

where
$$\bar{\beta}=\frac{\beta_1^0}{\beta_0^0}$$
 ---- Channel gain of interfering UE 1 Channel gain of UE 0

$$\frac{1}{\bar{\beta} + \frac{1}{\mathsf{SNR}_0}} = \frac{p\beta_0^0}{p\beta_0^0} + \sigma^2$$
Interference power Noise power



$$\left[\mathsf{SE}_0^{\mathrm{LoS}} \to \log_2\left(1 + \frac{1}{\bar{\beta}}\right) \quad \text{as} \ \ p \to \infty\right]$$



Ways to improve the Spectral Efficiency (SE)

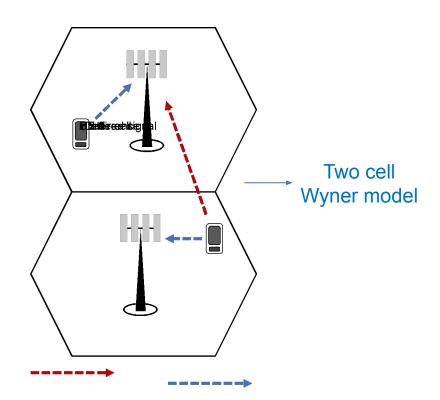


Array gain

- Increase the SE by using multiple transmitting antennas (M)
- Example: Non-Line-Of-Sight (NLOS) case. Two cells and two UEs

$$\mathsf{SE}_0^{\mathrm{NLoS}} \geq \log_2 \left(1 + rac{M-1}{ar{eta} + rac{1}{\mathsf{SNR}_0}}\right),$$

where
$$\bar{\beta} = \frac{\beta_1^0}{\beta_0^0}$$
 ---- Channel gain of interfering UE 1 Channel gain of UE 0



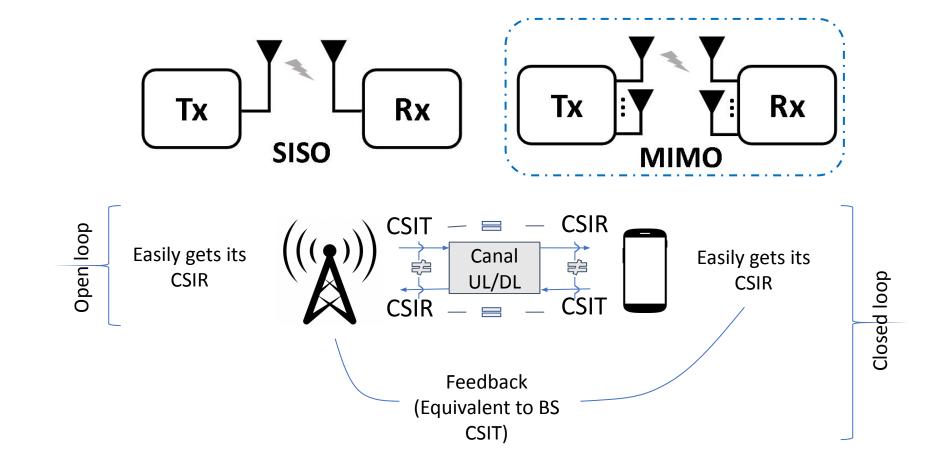


MIMO



Basics of multi antenna communication

Increase the SE by exploiting the array gain



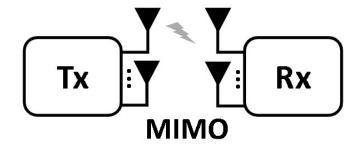


MIMO



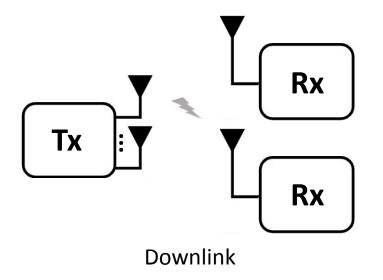
Single-user (SU) MIMO

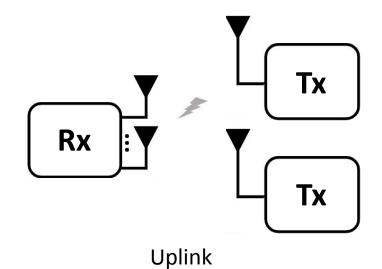
- There is only one UE and one BS (or AP)
- Both are equipped with multiple antennas



Multi-user (MU) MIMO

BSs are equipped with multiple antennas and serve multiple UEs (single or multi antenna)



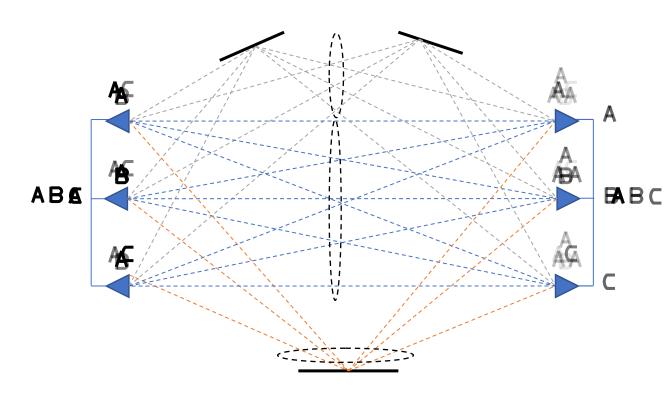




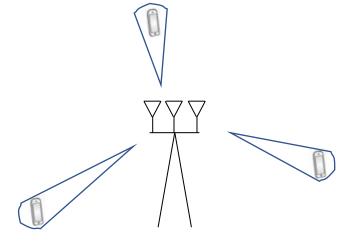
MIMO



MIMO operation types



- Diversity
 - Combiner
 - Space time code
- Multiplexing
 - Precoder
 - Combiner
- Multiple Acess
 - Precoder
 - Combiner

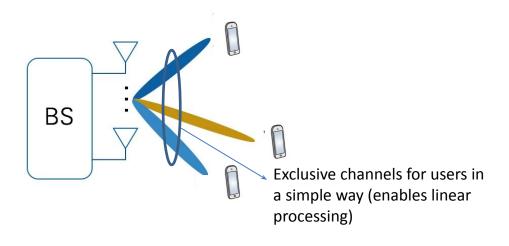






Basics of massive MIMO

- Base stations (BSs) equipped with several antennas (64, 128 or more)
- Number of users (K) smaller than M



Favorable propagation

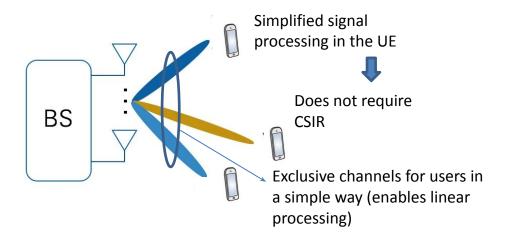
■ The channel directions become asymptotically orthogonal as $M \rightarrow \Box$





Basics of massive MIMO

- Base stations (BSs) equipped with several antennas (64, 128 or more)
- Number of users (K) smaller than M



Channel Hardening

A fast fading channel behaves almost deterministically due to spatial diversity

$$\frac{\|\mathbf{h}_{i}^{0}\|^{2}}{\mathbb{E}\{\|\mathbf{h}_{i}^{0}\|^{2}\}} = \frac{\|\mathbf{h}_{i}^{0}\|^{2}}{M\beta_{i}^{0}} \to 1 \quad \text{for} \quad M \to \infty$$





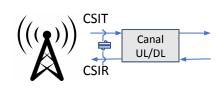
Basics of massive MIMO

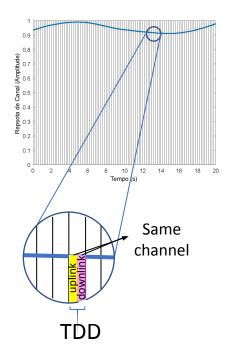
- Base stations (BSs) equipped with several antennas (64, 128 or more)
- Number of users (K) smaller than M

Simplified signal processing in the UE Does not require CSIR Exclusive channels for users in a simple way (enables linear processing)

TDD coherent operation

Reciprocal channels



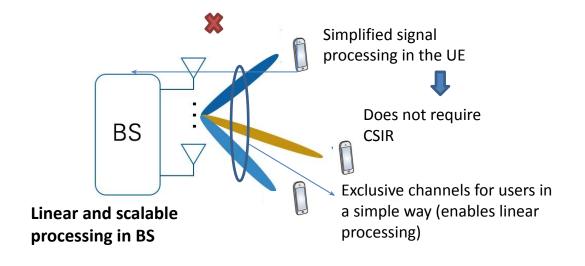




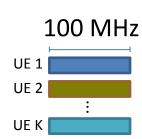


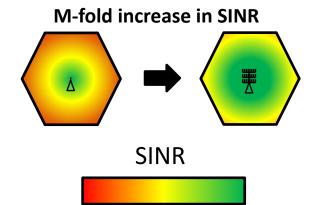
Basics of massive MIMO

- Base stations (BSs) equipped with several antennas (64, 128 or more)
- Number of users (K) smaller than M



K-fold increase in aggregate capacity









Distributed MIMO and Cell-free massive MIMO networks

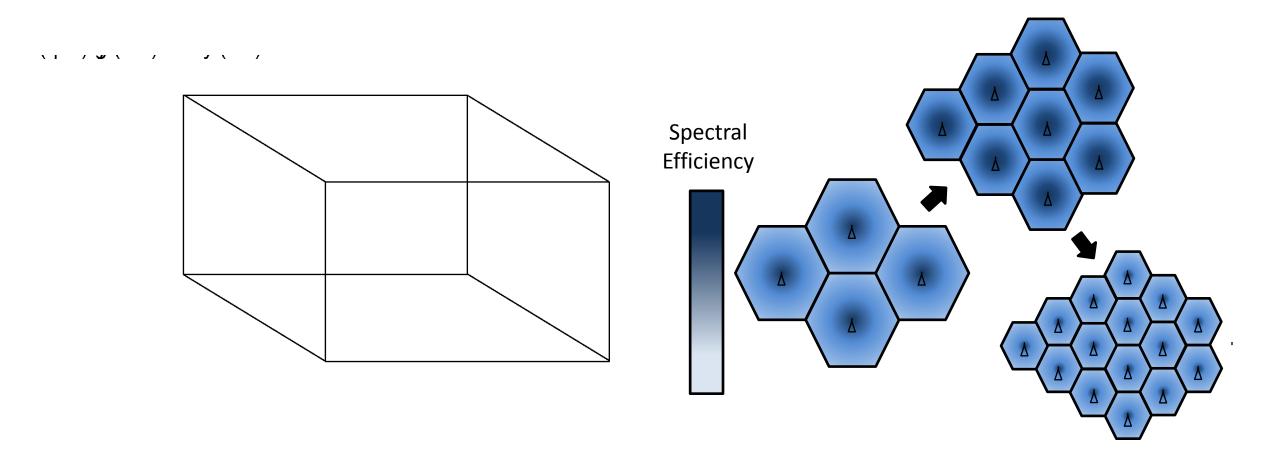






Densification effects of the cellular paradigm





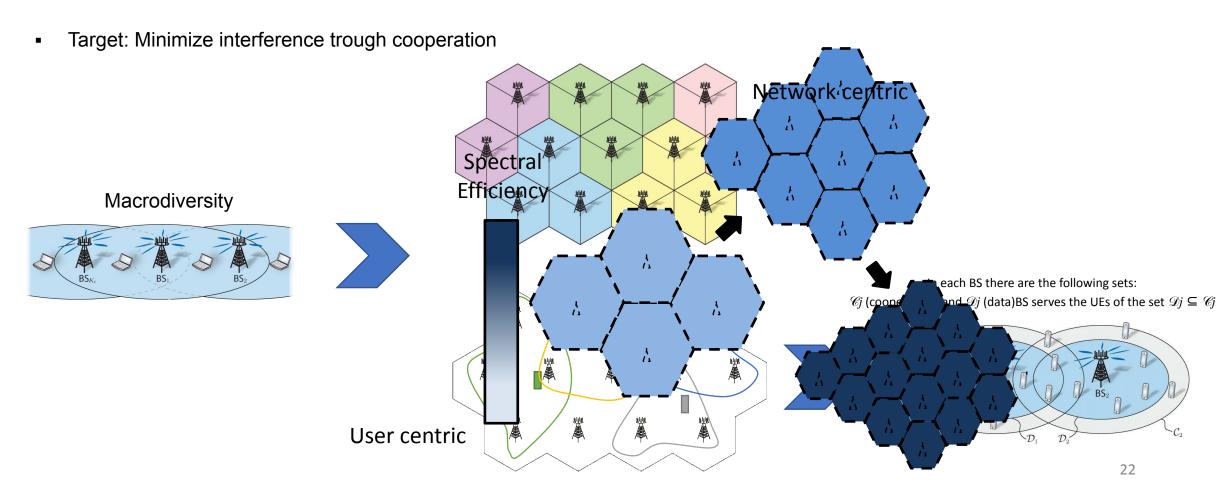


How to increase density without increasing interference



Distributed MIMO systems, network MIMO

A user is connected to more than one BS at the same time





Why is distributed MIMO not implemented yet



MIMO distribuído vs massive MIMO

Distributed MIMO Massive MIMO Formal definition (Standardization) TDD Signal processing definition Cell-Free



Cell-Free Massive MIMO Networks

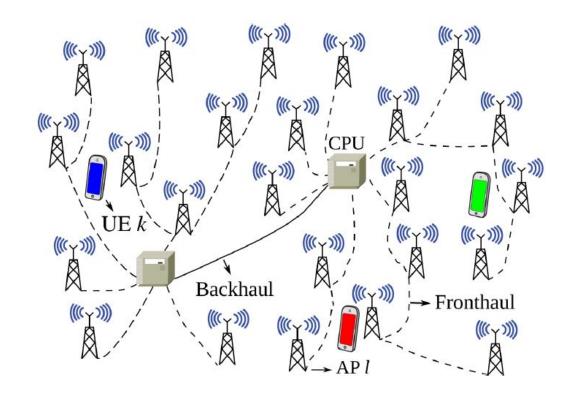


A particular case of distributed MIMO

- The antennas are spread out in the coverage area in terminals called access points (APs)
- Benefits of massive MIMO and distributed MIMO
- Better coverage probability

Canonical cell-free

- All APs serve all UEs
- Infeasible in terms of processing and data sharing





Evolution of Cell-Free Massive MIMO Networks



Timeline

- Distributed MIMO
- Canonical cell-free massive MIMO
- User-centric cell-free massive MIMO



Scalable user-centric cell free massive MIMO

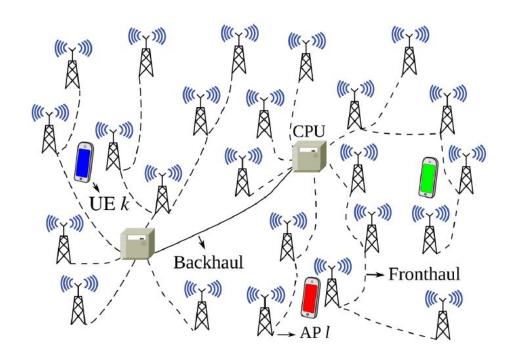


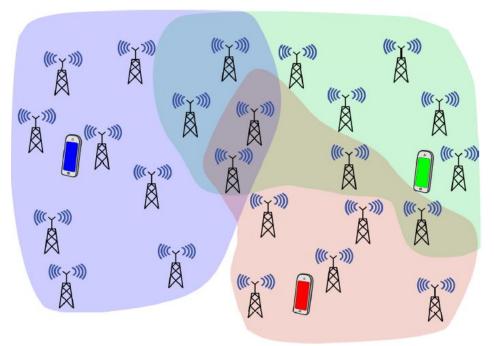
User-Centric Cell-Free Massive MIMO Networks



User centric approach

- The UEs connect to a subset of APs called AP cluster
- Reduction of processing demands and data sharing





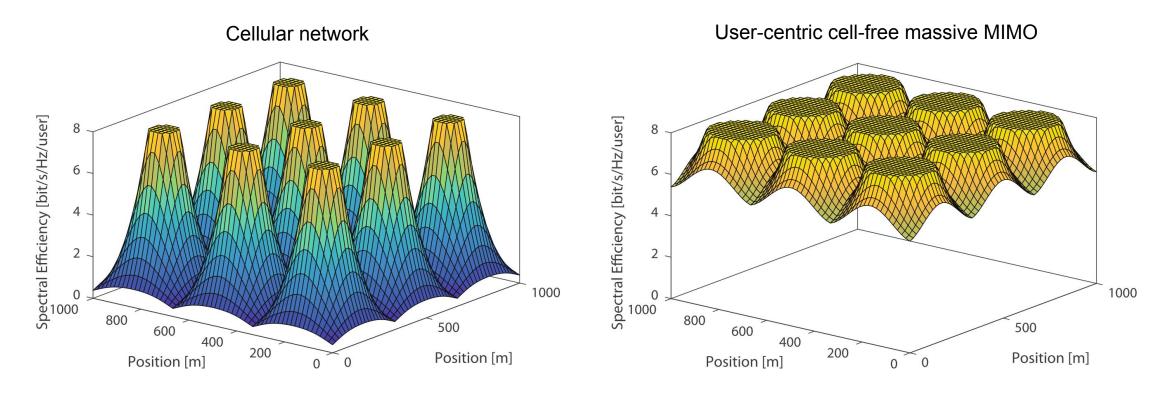


User-Centric Cell-Free Massive MIMO Networks



More uniform coverage than cell-based systems

- The SE of the 95% likely UEs is increased in user-centric cell-free massive MIMO systems
- Parameters setting: nine APs are deployed on a regular grid; UEs at different locations





Evolution of Cell-Free Massive MIMO Networks



Timeline

- Distributed MIMO
- Canonical cell-free massive MIMO
- User-centric cell-free massive MIMO
 - Also known as User-centric distributed massive MIMO
- Scalable user-centric cell free massive MIMO





Scalable user-centric cell free massive MIMO



Scalability

- User-centric cell-free massive MIMO systems are not scalable
- To determine whether a system is scalable or not, let $K \to \infty$ and check which of the following points are implementable:
 - 1) Signal processing for channel estimation
 - 2) Signal processing for data reception and transmission
 - 3) Fronthaul signaling for data and CSI sharing
 - 4) Power control optimization

Definition

- A cell-free massive MIMO systems is scalable if the four above-listed points have finite complexity and resource requirements for each AP as $K \to \infty$
- Necessary condition: limit the number of UEs that each AP can serve





Modelling a cell-free massive MIMO network



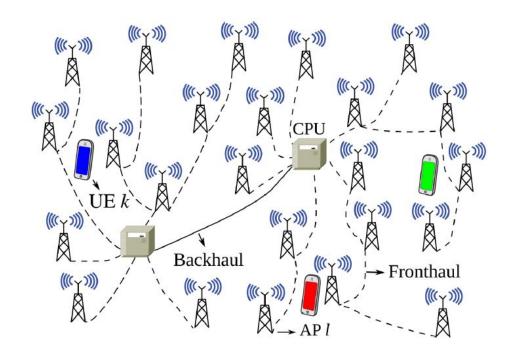


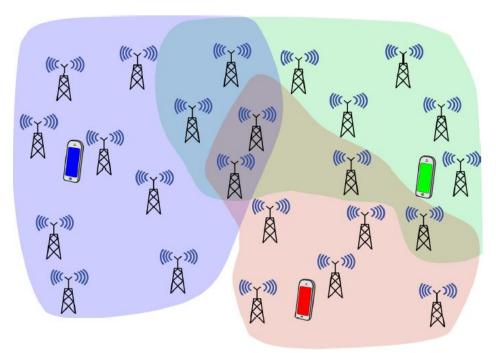




System model

- It is considered a cell-free massive MIMO network composed of L APs and K UEs
- Each AP is equipped with N antennas
- The total number of antennas in the network is M = NL





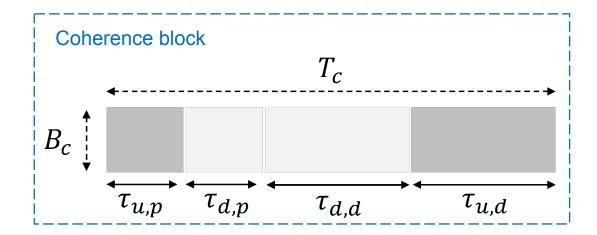


Time Division Duplexing (TDD) Protocol



Simplified frame structure and important concepts

- Coherence time (T_c) : the frequency response of a fading channel becomes flat (constant)
- Coherence bandwidth (B_c) : the bandwidth associated with the coherence time
- Coherence block (τ_c) : number of complex samples available for signal transmission



Number of samples per coherence block:

$$\tau = B_c T_c \longrightarrow T_c = \frac{\lambda}{4v} B_C = \frac{1}{2T_D}$$

$$\tau = \tau_{u,p} + \tau_{d,p} + \tau_{d,d} + \tau_{u,d}$$

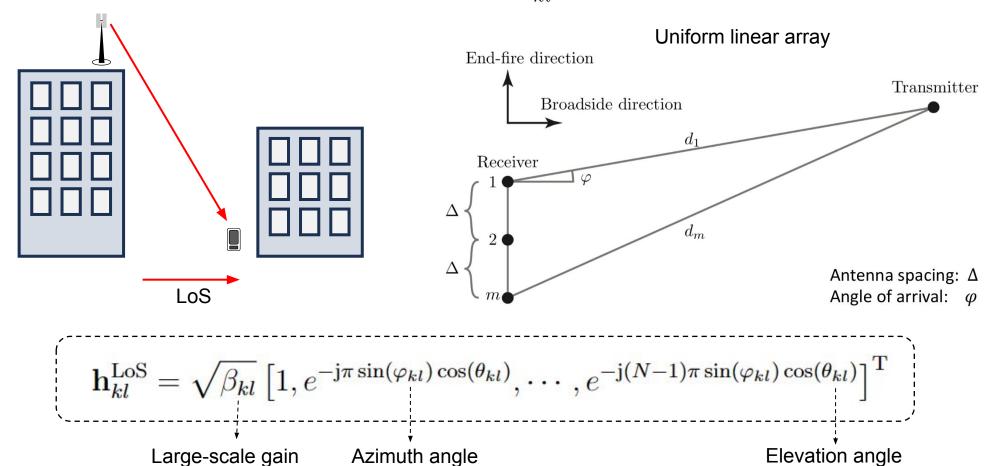
TDD can exploit reciprocity





System model: LoS component

- The channel vector between the AP / and UE k is denoted as $\mathbf{h}_{kl} \in \mathbb{C}^{N imes 1}$ and it can be calculated as

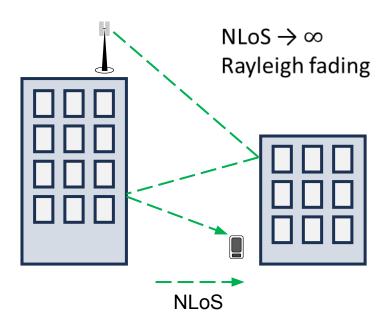




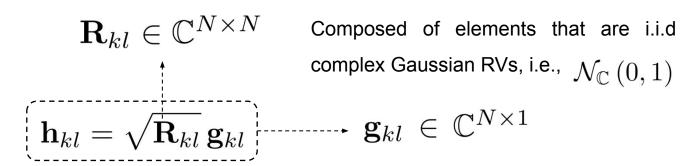


System model: NLoS component

The channel vector between the AP / and UE k is denoted as $\mathbf{h}_{kl} \in \mathbb{C}^{N imes 1}$ and it can be calculated as



Statistical covariance matrix



For all APs in the network, we have

$$\mathbf{h}_k = \begin{bmatrix} \mathbf{h}_{k1}^{\mathrm{T}}, ..., \mathbf{h}_{kL}^{\mathrm{T}} \end{bmatrix}^{\mathrm{T}} \in \mathbb{C}^{M \times 1}$$
 $\mathbf{R}_k = \mathrm{diag}\left(\mathbf{R}_{k1}, ..., \mathbf{R}_{kL}\right) \in \mathbb{C}^{M \times M}$

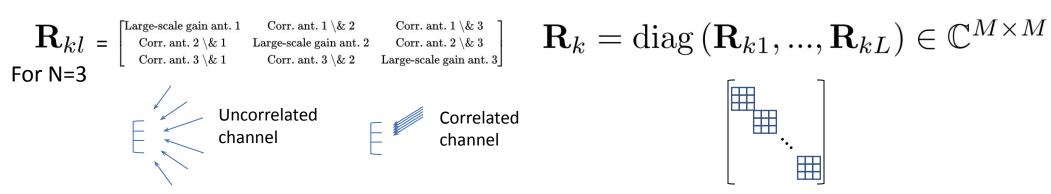
Collective vector



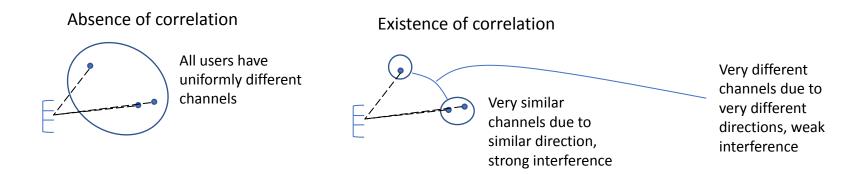


System model: Correlation matrix

 Matrix describing the macroscopic effects of propagation, including antenna gains and radiation patterns at the transmitter and receiver.



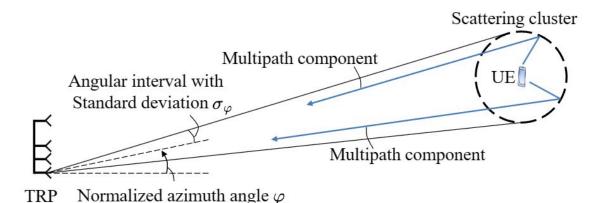
The presence of correlation in the channel can favor or hinder interference cancellation







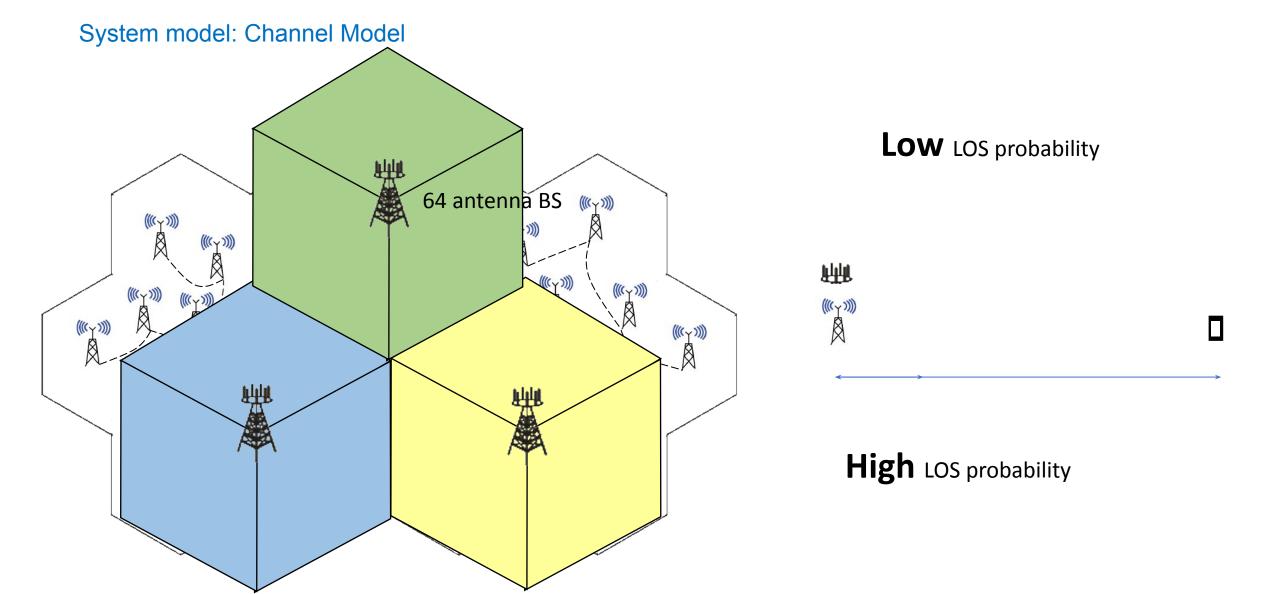
System model: Correlation matrix



$$\begin{split} [\mathbf{R}]_{l,m} &= \beta \int e^{2\pi \mathrm{j} d_{\mathrm{H}}(l-m)\sin(\bar{\varphi})} f(\bar{\varphi}) d\bar{\varphi} & \bar{\varphi} = \varphi + \delta \\ [\mathbf{R}]_{m\ell} &= \beta \int \int e^{2\pi \mathrm{j} d_{\mathrm{H}}(l-m)\sin(\bar{\varphi})} f(\varphi,\theta) d\varphi d\theta & f(\bar{\varphi},\bar{\theta}) = \frac{1}{2-l} e^{-\frac{(\bar{\varphi}-\varphi)^2}{2\sigma_{\varphi}^2}} e^{-\frac{(\bar{\theta}-\theta)^2}{2\sigma_{\theta}^2}} \\ \delta &\sim \mathcal{N}(0,\sigma_{\varphi}^2) & \delta \sim U[-\sqrt{3}\sigma_{\varphi},\sqrt{3}\sigma_{\varphi}] & \delta \sim \mathrm{Lap}(0,\sigma_{\varphi}/\sqrt{2}) \end{split}$$





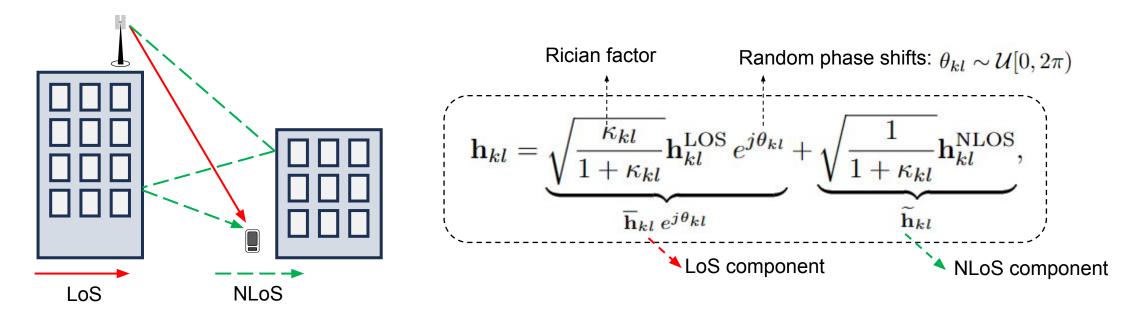






System model: Channel Model

- The channel vector between the AP / and UE k is denoted as $\mathbf{h}_{kl} \in \mathbb{C}^{N imes 1}$ and it can be calculated as



Rician factor is the power ratio between LoS and NLoS component

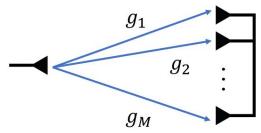
$$\kappa_{kl}=10^{13-0.03d_{lk}}$$
 Distance $\kappa_{kl}=rac{p_{ ext{LoS}}(d_{lk})}{1-p_{ ext{LoS}}(d_{lk})}$ LoS probability



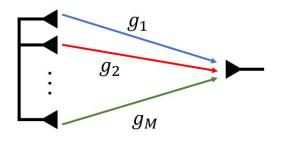


System model: channel estimation

The channel vector is estimated by transmitting pilots



One pilot: Estimate all coefficients



M pilots to estimate all coefficients

Pilots are known signals designed to be mutually orthogonal

$$\phi_{t_1}^{ ext{H}} \phi_{t_2} = egin{cases} \sqrt{ au_p}, & ext{if } t_1 = t_2 \ 0, & ext{if } t_1
eq t_2. \end{cases}$$

Example: [0 0 0 1], [0 0 1 0], [0 1 0 0], [1 0 0 0]

Scales with the

number of transmitters





System model: channel estimation

AP / receives the signal of all users

$$\mathbf{y}_l^{ ext{pilot}} = \sum_{k=1}^K \sqrt{\eta_k} \mathbf{h}_{kl} oldsymbol{\phi}_{t_k} + \mathbf{n}_l$$
 Noise

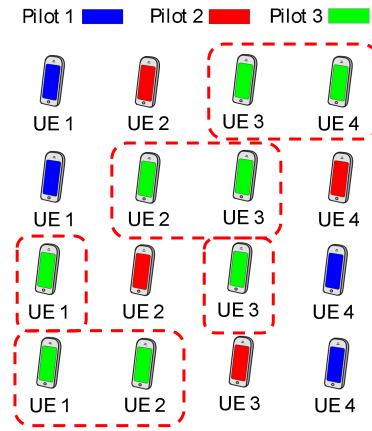
To estimate the channel of UE k, AP l correlates with its pilot

$$\mathbf{y}_{t_k l}^{ ext{pilot}} = \phi_{t_k}^{ ext{H}} \mathbf{y}_l^{ ext{pilot}} = \sum_{i \in \mathcal{P}_k} \sqrt{ au_p \eta_i} \mathbf{h}_{il} + \mathbf{n}_{t_k l},$$
 All UEs with pilot index t_k

Pilot interference happens among users with the same pilot

$$\mathbb{E}\left\{\widehat{\mathbf{h}}_{kl}\widehat{\mathbf{h}}_{il}^{\mathrm{H}}\right\} = \sqrt{\eta_{k}\eta_{i}}\tau_{p}\mathbf{R}_{kl}\boldsymbol{\Psi}_{t_{k}l}^{-1}\mathbf{R}_{il}, \quad i\in\mathcal{P}_{k}. \qquad \text{UE1} \quad \text{UE2} \quad \text{UE3}$$

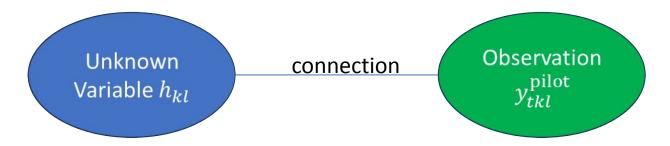
$$\mathbf{R}_{kl} = \mathbb{E}\{\mathbf{h}_{kl}\mathbf{h}_{kl}^{\mathrm{H}}\} = (\overline{\mathbf{h}}_{kl}\overline{\mathbf{h}}_{kl}^{\mathrm{H}} + \widetilde{\mathbf{R}}_{kl}) \qquad \boldsymbol{\Psi}_{t_{k}l} = \mathbb{E}\{(\mathbf{y}_{t_{k}l}^{\mathrm{pilot}})(\mathbf{y}_{t_{k}l}^{\mathrm{pilot}})^{\mathrm{H}}\} = \sum_{i\in\mathcal{P}_{k}}\eta_{i}\tau_{up}\mathbf{R}_{il} + \sigma_{\mathrm{ul}}^{2}\mathbf{I}_{N}$$







System model: channel estimation



Least-square (LS) estimation

$$\widehat{\mathbf{h}}_{kl} = rac{1}{\sqrt{\eta_k au_p}} \mathbf{y}^{ ext{pilot}}_{t_k l}$$

Minimizes
$$\left\|\mathbf{y}_{t_k l}^{\mathrm{pilot}} - \sqrt{p_k au_p} \mathbf{h}_{k l} \right\|^2$$

No need for channel statistics

Minimum mean square estimation (MMSE)

$$\widehat{\mathbf{h}}_{kl} = \sqrt{\tau_p \eta_k} \mathbf{R}_{kl} \mathbf{\Psi}_{tl}^{-1} \mathbf{y}_{tl}^{\mathrm{pilot}}$$

Minimizes
$$\mathbb{E}\{\|\mathbf{h}_{kl}\!-\!\widehat{\mathbf{h}}_{kl}\|^2\}$$

Needs channel statistics

Known channel and noise distribution



AP cluster matrix

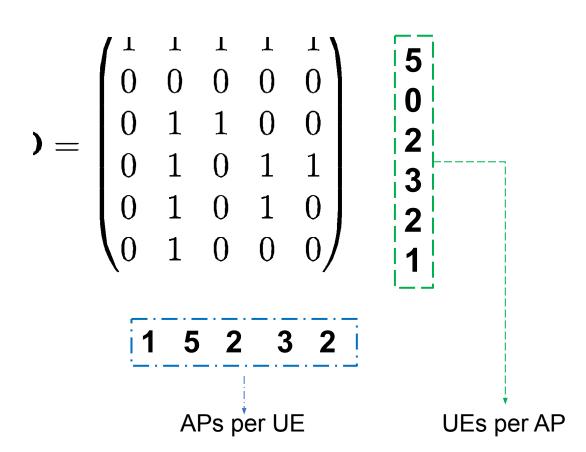


Binary matrix indicating the APs serving the UE

- Dimension: L x K.
- In this example, L = 6 and K = 5

The matrix **D** also indicates the subset of UEs that each AP serves

• The number of UEs that each AP serves is limited in a scalable system

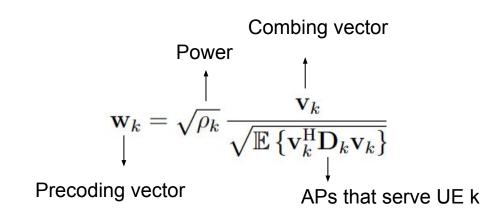


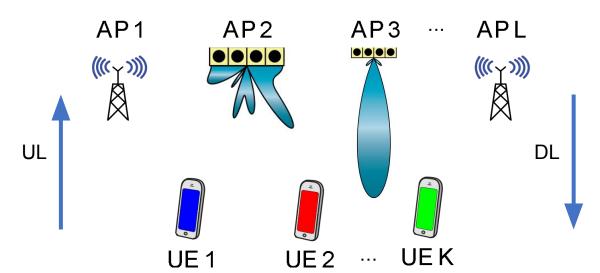




System model: data transmission

- UL phase:
 - UEs send data to the APs
 - Combining vectors are computed based on channel estimates
- DL phase:
 - APs send data to the UEs
 - Precoding vectors are computed based on combining vectors
- Combining and precoding are signal-processing techniques used to improve the desired signal and mitigate interference



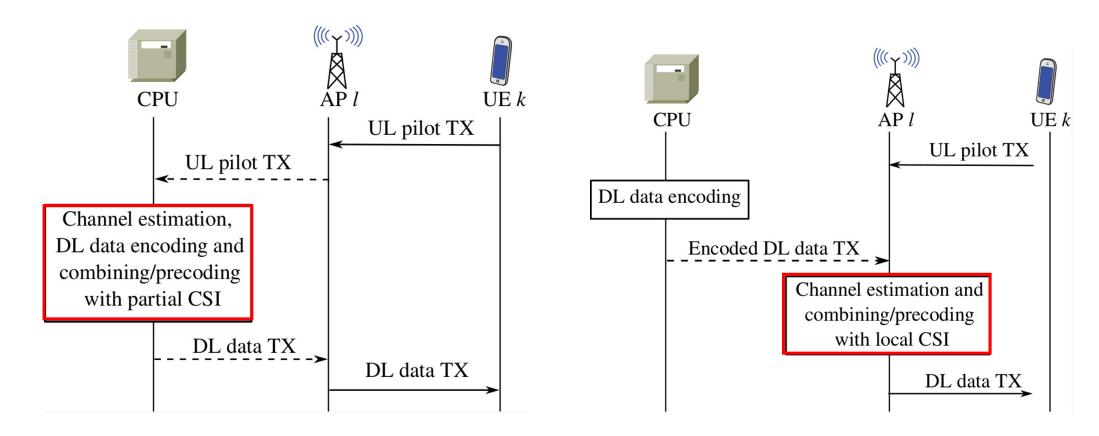






System model: data transmission

Combining and precoding can be performed at CPUs or APs

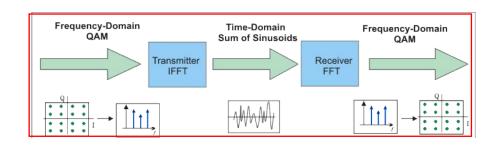


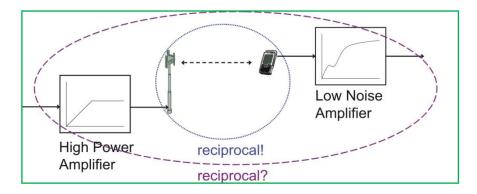




System model: network implementation

- Distribution of the processing functions in CPUs and RUs for centralized and distributed implementations
- Each task has an associated processing computational complexity and power consumption





Centralized implementation

CPUs

Higher layer network

Higher layer control

Channel coding

Mapping/demapping

OFDM

modulation/demodulation

Channel estimation

Combiner computation

Precoding/combining

Reciprocity calibration



APs

FFT/IFFT

Baseband filtering

Distributed implementation

CPUs

Higher layer network

Higher layer control

Channel coding

Mapping/demapping

OFDM

modulation/demodulation



APs

Channel estimation

Combiner computation

Precoding/combining

Reciprocity calibration

FFT/IFFT

Baseband filtering





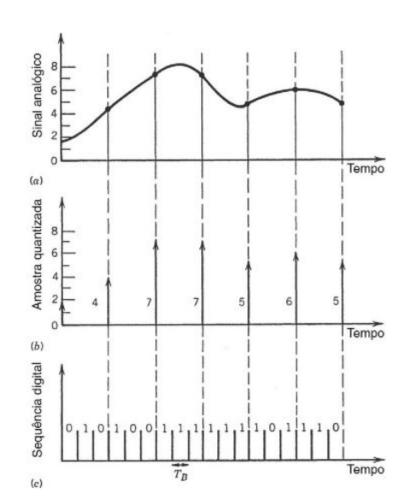
System model: distortions due to fronthaul limitations

Bussgang Decomposition

$$\hat{\mathbf{y}} = \mathcal{Q}_{\theta}(\mathbf{y}) \approx \alpha_{\theta} \mathbf{y} + \tilde{\mathbf{q}}_{\theta},$$

Gaussian

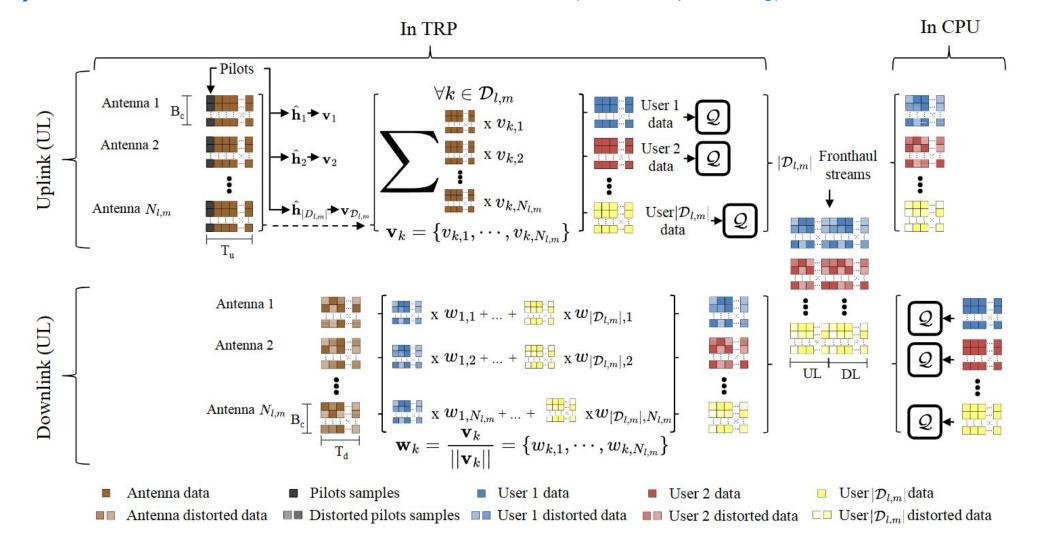
| Bits per sample | Distortion rate | Correlation coefficient (α_{θ}) |
|-----------------|------------------|--|
| 1 | 0.3634 | 0.6366 |
| 2 | 0.1175 | 0.8825 |
| 3 | 0.03454 | 0.96546 |
| 4 | 0.009497 | 0.990503 |
| 5 | 0.002499 | 0.997501 |
| > 5 | $1-lpha_{	heta}$ | $1 - \frac{\pi\sqrt{3}}{2}2^{-2(\text{bits})}$ |







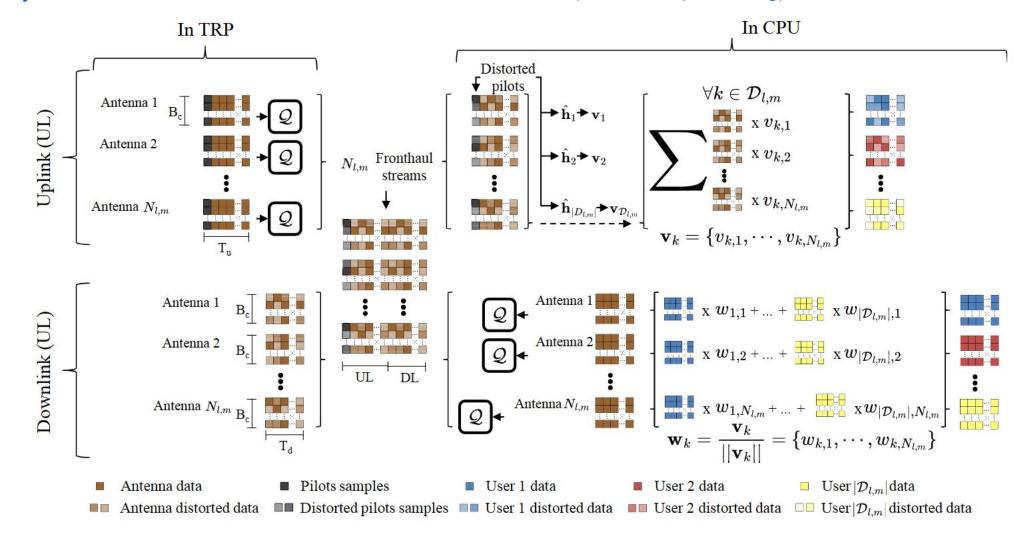
System model: distortions due to fronthaul limitations (distributed processing)







System model: distortions due to fronthaul limitations (centralized processing)





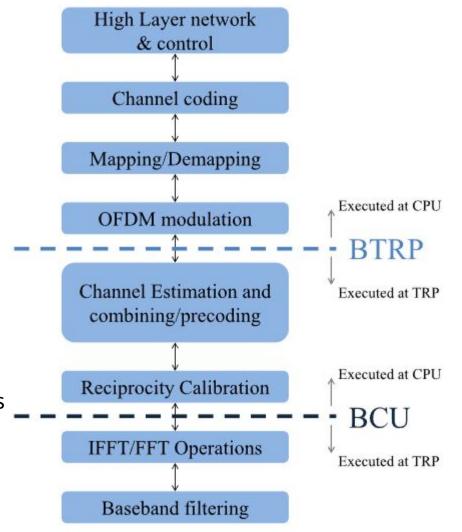


System model: fronthaul bitrate

$$F_{l,m,t} = \begin{cases} 2B \left(1 - \frac{\tau_p}{\tau_c}\right) \sum_k b_{l,k}^{\text{data}}, & \text{for BTRP} \\ 2N B \left[\left(1 - \frac{\tau_p}{\tau_c}\right) b_l^{\text{data}} + \frac{\tau_p}{\tau_c} b_l^{\text{pil}} \right], & \text{for BCPU} \end{cases}$$

Example for a maximum of 10 GB of capacity, 10 TRPs (4 antennas), 8 users and 20 pilots, B=100 MHz, and a coherence block with 200 samples:

$$10*10^9/(2*100*10^6*(1-20/200)*8) = b_{l,k}^{data}$$
 $6,94 = b_{l,k}^{data}$
7 users will have 7 bits
 $1 \text{ user will have 6 bits}$
 $10*10^9/(2*4*100*10^6)=12.5$
 $b_l^{\text{pil}}=17 \text{bits}$
 $b_l^{\text{data}}=12 \text{ bits}$
 $12.5 = \left(1-\frac{\tau_p}{\tau_c}\right) b_l^{\text{data}} + \frac{\tau_p}{\tau_c} b_l^{\text{pil}}$







System model: centralized combiners

P-MMSE: finds the trade-off of maximizing the desired signal and mitigating interference

$$\mathbf{v}_k^{\mathrm{P-MMSE}} = p_k \left(\sum_{i \in \mathcal{S}_k} p_i \mathbf{D}_k \widehat{\mathbf{h}}_i \widehat{\mathbf{h}}_i^{\mathrm{H}} \mathbf{D}_k + \mathbf{Z}_{\mathcal{S}_k} + \sigma_{\mathrm{ul}}^2 \mathbf{I}_{LN} \right)^{-1} \mathbf{D}_k \widehat{\mathbf{h}}_k$$
 UEs that most interfere with UE k
$$\mathbf{Z}_{\mathcal{S}_k} = \sum_{i \in \mathcal{S}_k} p_i \mathbf{D}_k \mathbf{C}_i \mathbf{D}_k$$
 Correlation matrix of the estimation error

P-RZF: interference between the UEs tends to zero

$$\mathbf{v}_k^{\mathrm{P-RZF}} = \left[\mathbf{D}_k \widehat{\mathbf{H}}_{\mathcal{S}_k} \left(\widehat{\mathbf{H}}_{\mathcal{S}_k}^{\mathrm{H}} \mathbf{D}_k \widehat{\mathbf{H}}_{\mathcal{S}_k} + \sigma_{\mathrm{ul}}^2 \mathbf{P}_{\mathcal{S}_k}^{-1} \right)^{-1} \right]_{:,1}$$
Neglects $\mathbf{Z}_{\mathcal{S}_k}$
Regularized pseud-inverse





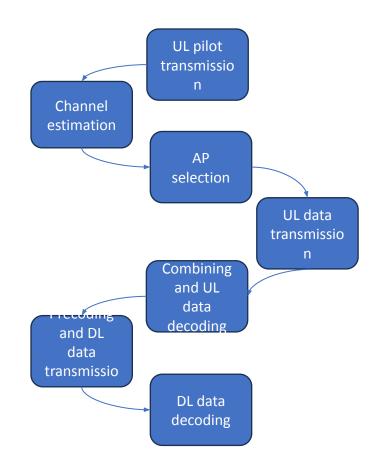
System model: distributed combiners

LP-MMSE: same as P-MMSE, but uses only local channel estimates

$$\mathbf{v}_{kl}^{\mathrm{LP-MMSE}} = p_k \left(\sum_{i \in \mathcal{D}_l} p_i \left(\widehat{\mathbf{h}}_{il} \widehat{\mathbf{h}}_{il}^{\mathrm{H}} + \mathbf{C}_{il} \right) + \sigma_{\mathrm{ul}}^2 \mathbf{I}_N \right)^{-1} \mathbf{D}_{kl} \widehat{\mathbf{h}}_{kl}.$$
 UEs served by AP /

MR: maximizes the desired signal

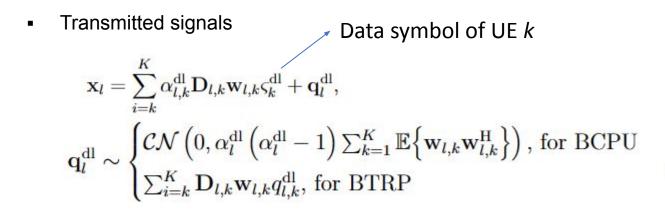
$$\mathbf{v}_{kl}^{\mathrm{MR}} = \widehat{\mathbf{h}}_{kl}.$$



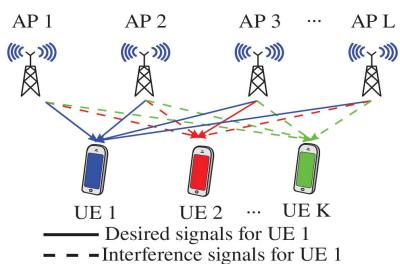




System model: DL data transmission



$$q_{l,k}^{\mathrm{ul}} \sim \mathcal{CN}\left(0, \alpha_{l,k} \left(\alpha_{l,k} - 1\right)\right)$$



Received signals

$$egin{aligned} y_k^{ ext{dl}} &= \sum_{l=1}^L \sum_{m=1}^{M_l} (\mathbf{h}_{l,k})^H \mathbf{x}_l + n_k^{ ext{dl}}, \ y_k^{ ext{dl}} &= \sum_{l=1}^L lpha_{l,k}^{ ext{dl}} \mathbf{h}_{l,k}^H \mathbf{D}_{l,k} \mathbf{w}_{l,k}^{ ext{dl}} s_k^{ ext{dl}} \ &+ \sum_{\substack{i=1 \ i
eq k}}^K \sum_{l=1}^L \sum_{m=1}^{M_l} lpha_{l,i}^{ ext{ul}} \mathbf{h}_{l,k}^H \mathbf{D}_{l,i} \mathbf{w}_{l,i}^{ ext{dl}} s_i^{ ext{dl}} \ &+ \sum_{\substack{i=1 \ i
eq k}}^L \mathbf{h}_{l,k}^H \mathbf{q}_l \ . \end{aligned}$$
Inter-User Interference
$$+ \underbrace{n_k^{ ext{dl}}}_{\text{Receiver Noise}} + \underbrace{\sum_{l=1}^L \mathbf{h}_{l,k}^H \mathbf{q}_l}_{\text{Quantization Noise}} \ . \end{aligned}$$





System model: DL spectral efficiency

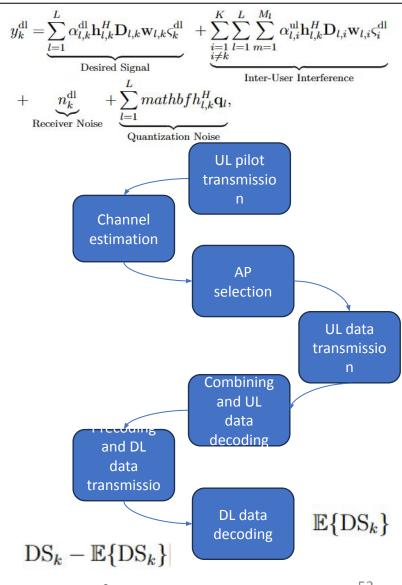
If UEs have only statistical CSI knowledge, i.e., E {αkk}

Pre-log factor

$$R_k^{\text{UatF}} = \phi B \log_2(1 + \underline{\text{SINR}}_k), \quad \text{SE}_k = \phi \log_2(1 + \underline{\text{SINR}}_k),$$

$$\frac{|\mathbb{E}\{\mathrm{DS}_k\}|^2}{\mathbb{E}\{|\mathrm{DS}_k-\mathbb{E}\{\mathrm{DS}_k\}|^2\}+\mathbb{E}\{|\mathrm{IS}_k|^2\}+\mathbb{E}\{|\mathrm{QN}_k|^2\}+\sigma_{\mathrm{ul}|\mathrm{dl}}^2},}$$
 Noise power

Known as hardening bound

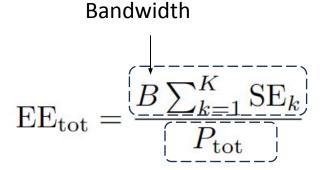




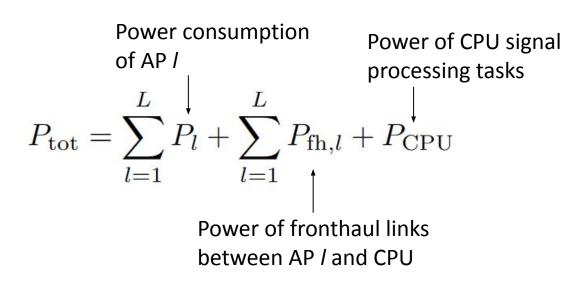


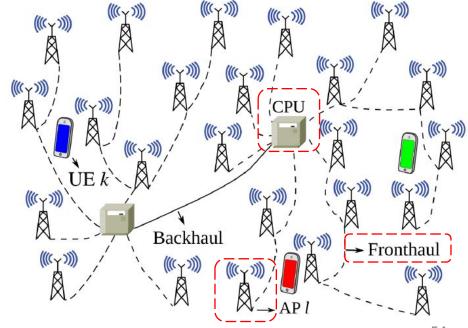
System model: Energy efficiency











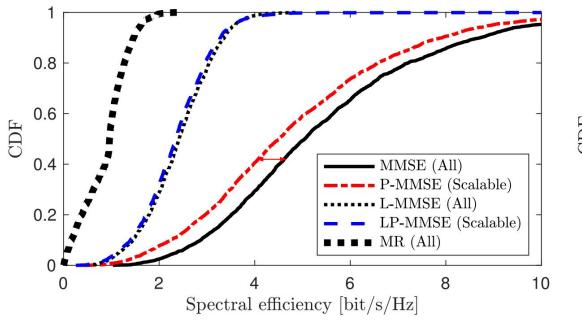


Simulation results



Scalable cell-free system vs canonical cell-free

- Negligible losses in SE
- The distributed implementation (LP-MMSE) is almost not affected



0.8

0.6

0.4

0.2

MMSE (All)

P-MMSE (Scalable)

L-MMSE (All)

LP-MMSE (Scalable)

MR (All)

Spectral efficiency [bit/s/Hz]

Parameters setting: L = 400 APs, K = 100 UEs, N = 1

Parameters setting: L = 100 APs, K = 100 UEs, N = 4



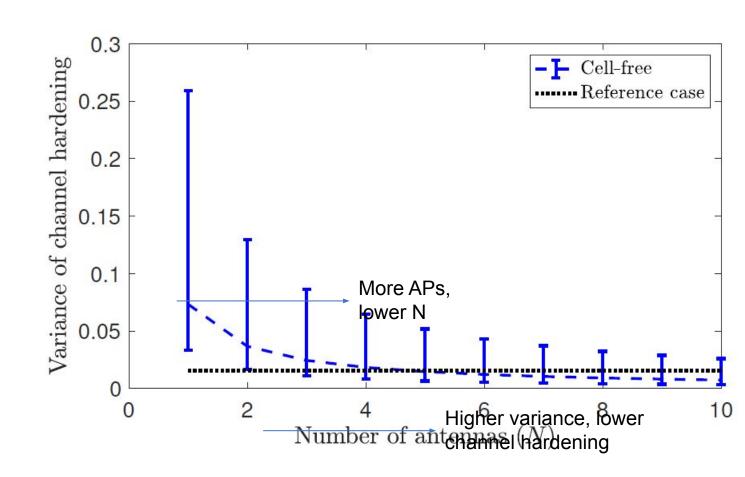


System model: channel hardening

$$CH_k = rac{\mathrm{VAR}\left(\eta_{k,k}
ight)}{\left(\mathbb{E}\left[\eta_{k,k}
ight]
ight)^2}$$

$$\eta_{k,k'} = \sum_{l=1}^{L} \alpha_{l,k}^{DL} \boldsymbol{h}_{l,k}^{H} \boldsymbol{D}_{l,k'} \boldsymbol{w}_{l,k'}, \quad k \in \{1,2,\cdots,K\}$$

Reference case: cellular system, i.i.d. Rayleigh fading \rightarrow 1/L





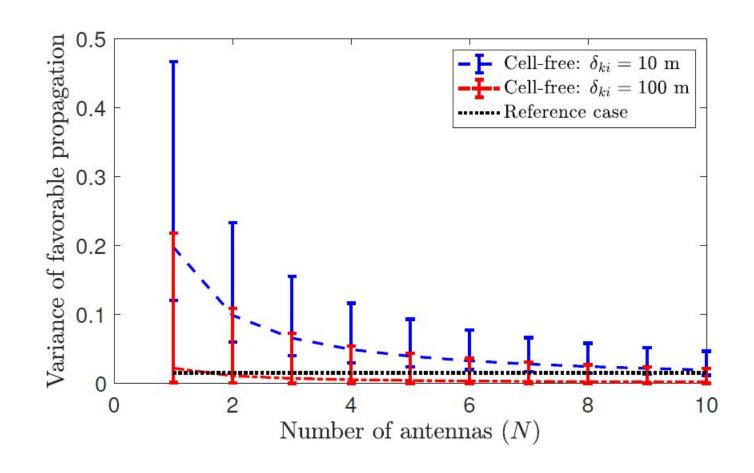


System model: favorable propagation

Interfering effective channel
$$PP_k = rac{ ext{VAR}(\eta_{k,k'})}{\left(\mathbb{E}[\eta_{k,k}]
ight)^2}, \ orall k'
eq k$$

$$\eta_{k,k'} = \sum_{l=1}^{L} \alpha_{l,k}^{DL} \boldsymbol{h}_{l,k}^{H} \boldsymbol{D}_{l,k'} \boldsymbol{w}_{l,k'}, \quad k \in \{1, 2, \dots, K\}$$

Reference case: cellular system, i.i.d. Rayleigh fading \rightarrow 1/L







A comparison among AP selection strategies





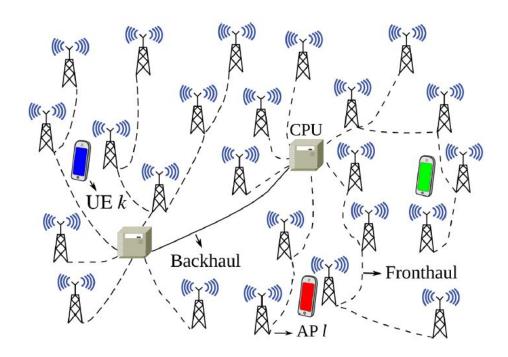


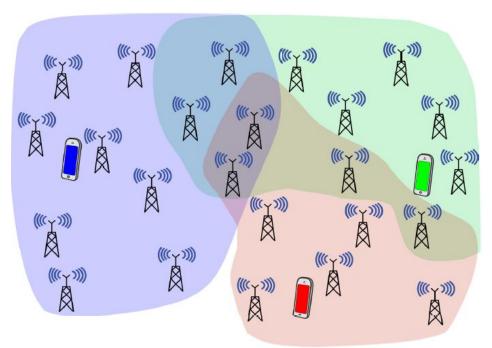
AP selection (AP clustering)



AP selection schemes (common approaches)

- Canonical cell-free (CF) massive MIMO
- User-centric clustering (UCC)
- Largest-large-scale-fading-based (LSFB)





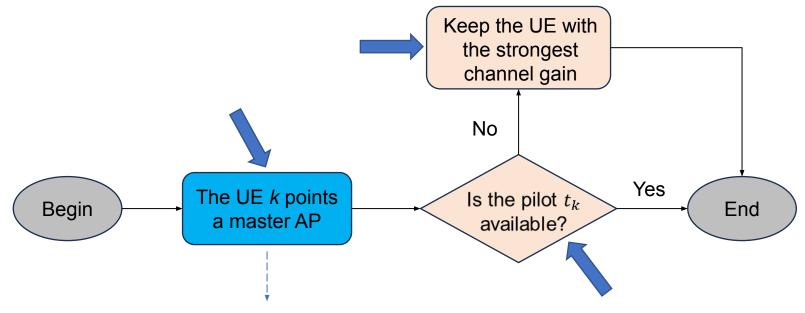


Scalable cell-free (CF)



Relates AP selection process to pilot assignment

• In this scheme each AP can serve up to $\tau_{u,p}$ UEs



- Assigns a pilot t_k
- Informs the other APs that the UE k has entered the network

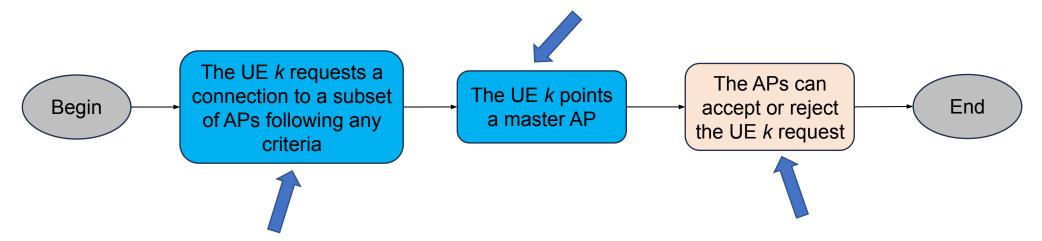


Matched Decision



A general framework

- The AP selection considers the most suitable connections for both UEs and APs
- It can be adjusted according to different criteria



It can behave like the previous AP selection schemes, but providing scalability and guaranteeing connection

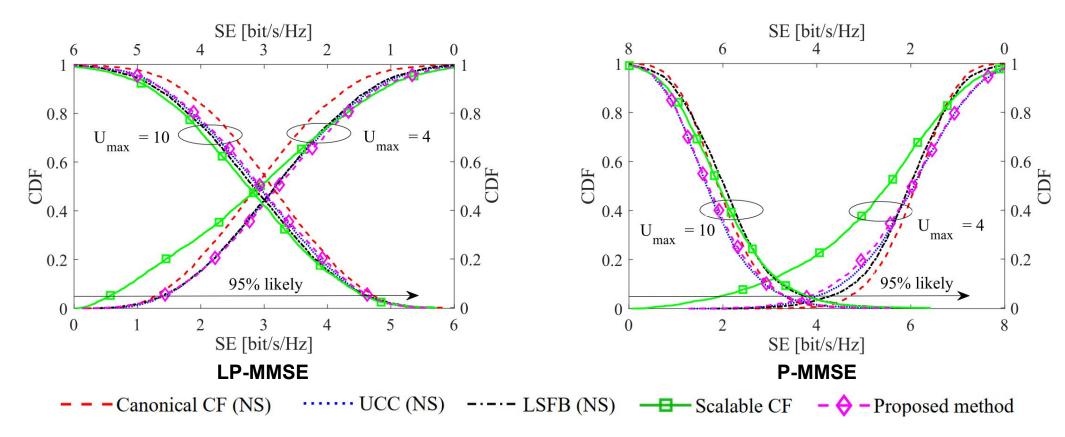


AP selection (AP clustering)



A comparison, downlink (DL) spectral efficiency (SE)

- Parameters setting: L = 100 APs, K = 25 UEs, N = 1
- U_{max} denotes the maximum number of UEs that each AP can serve



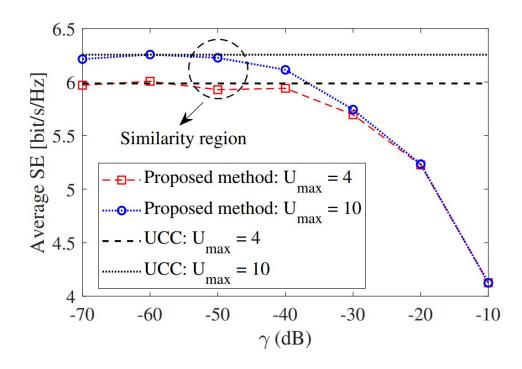


AP selection (AP clustering)



A comparison, downlink (DL) spectral efficiency (SE)

 The matched decision can behave like other AP selection strategies while affording scalability and ensuring connection of the UE with the network



Time complexity Ar serection method Complexity

| UCC | $\mathcal{O}(LK\log K)$ |
|------------------|---|
| LSFB | $\mathcal{O}(L \log L)$ |
| Scalable CF | $\mathcal{O}\left(L	au_{p} ight)$ |
| Matched Decision | $\mathcal{O}\left(L\hat{U}_{	ext{max}} ight)$ |



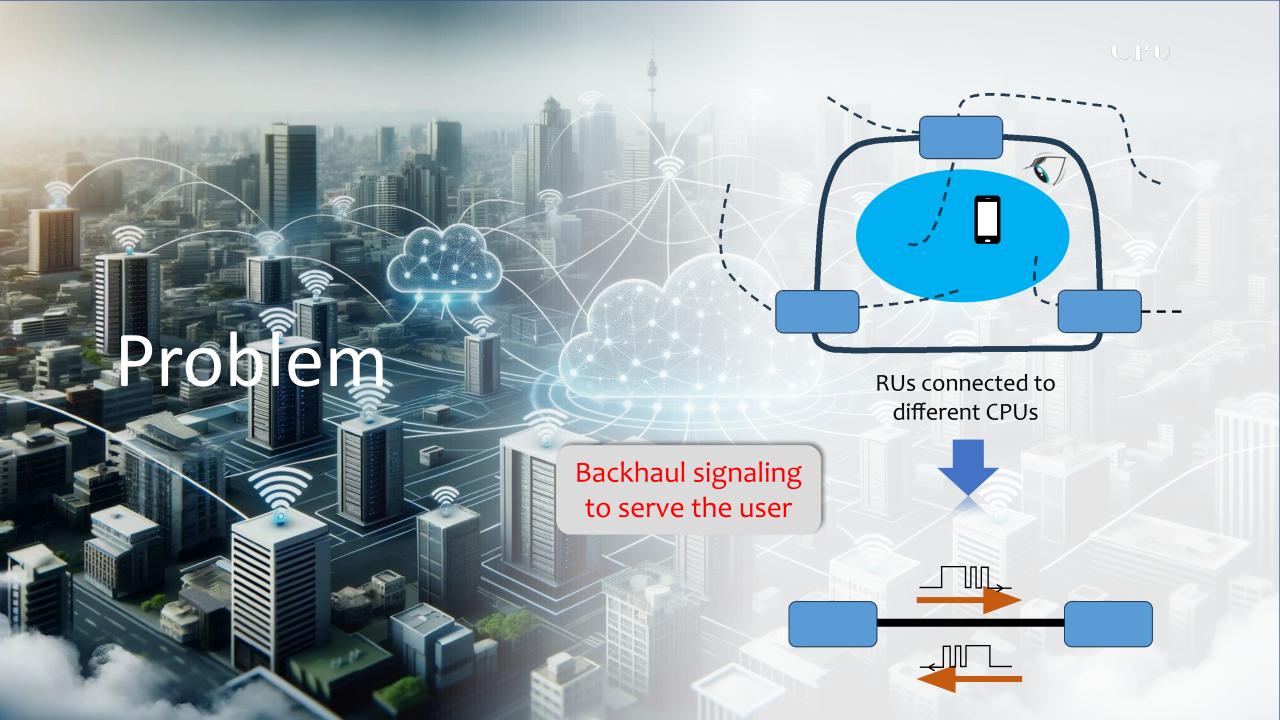


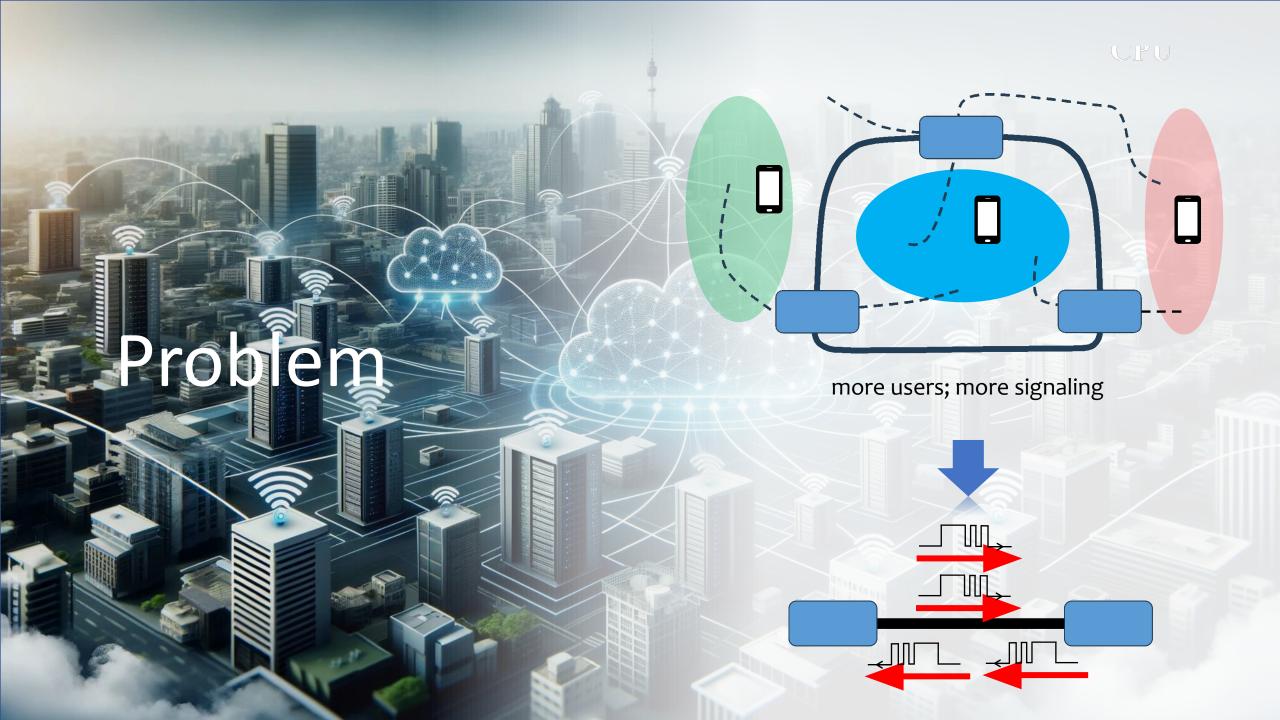
Some researches regarding user-centric cell-free massive MIMO systems



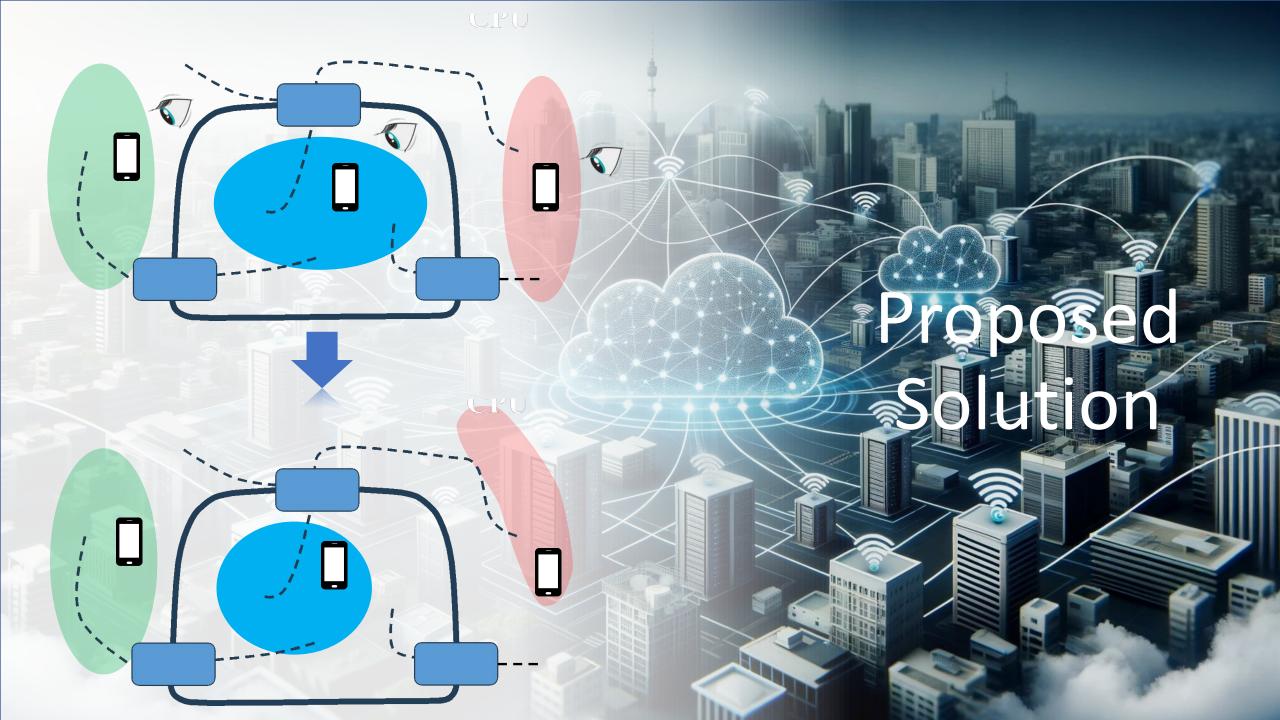


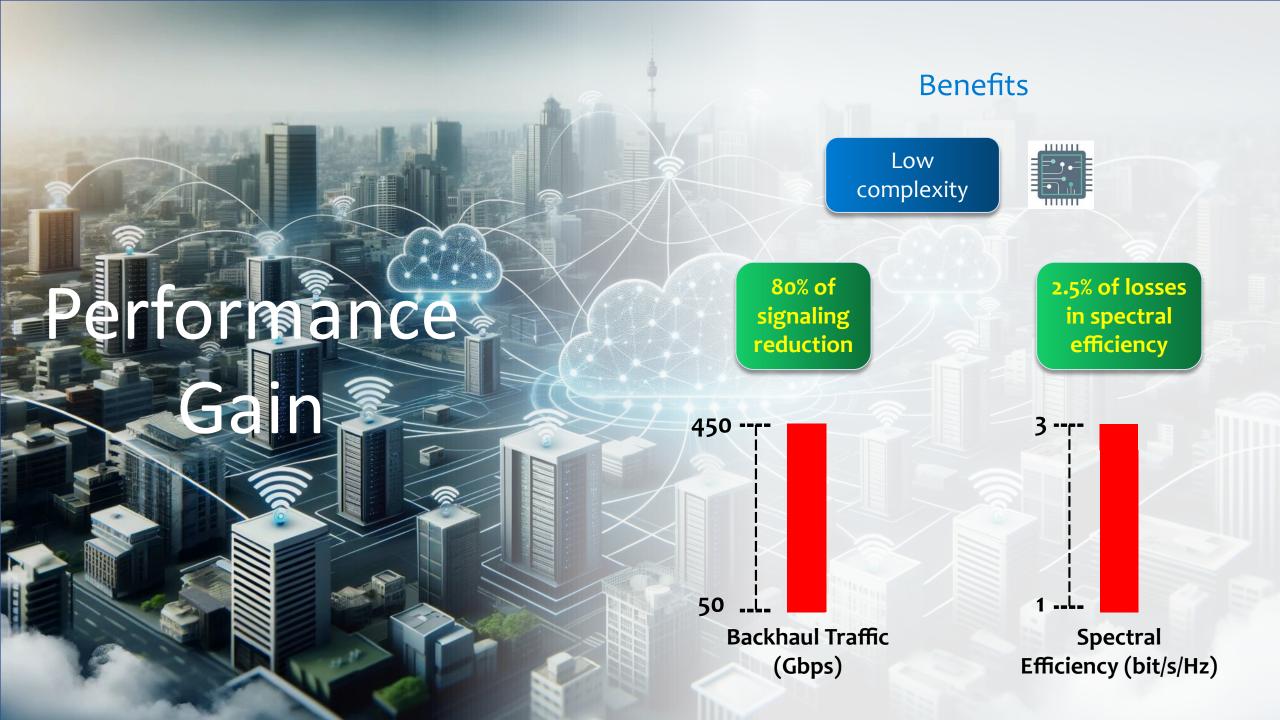


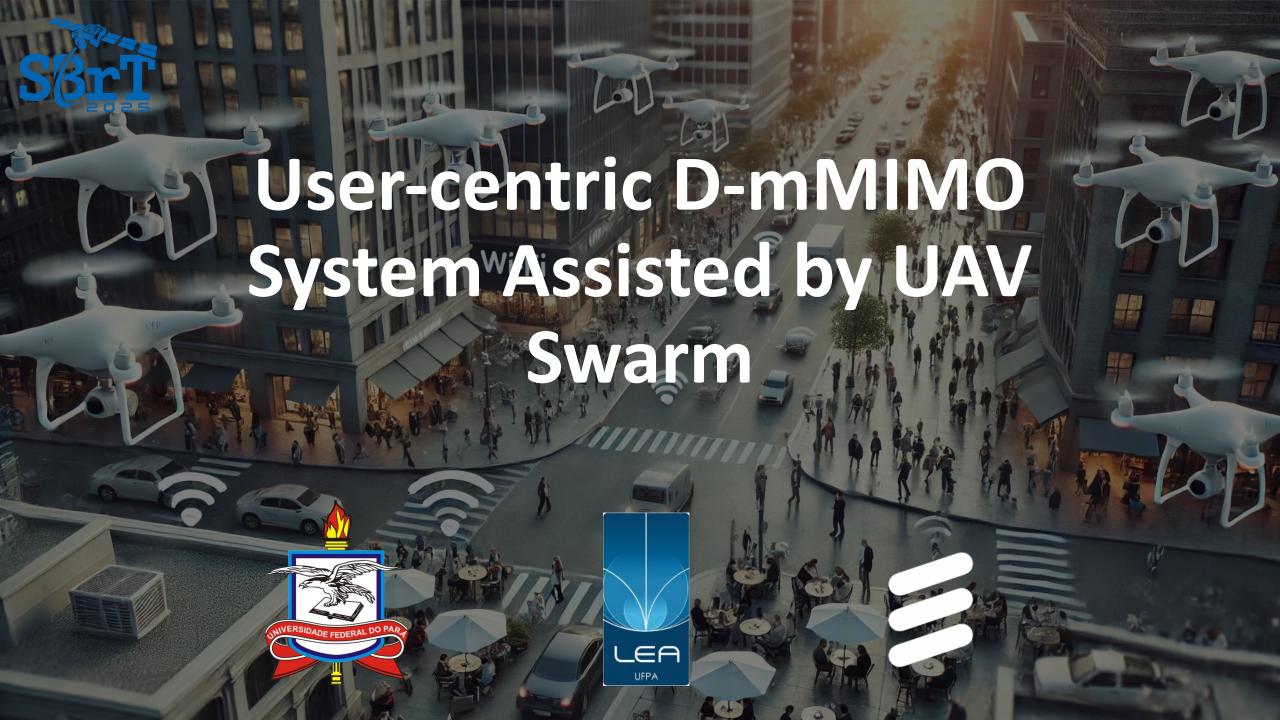








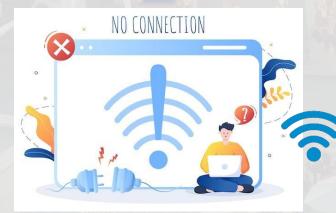






Natural disasters disrupt power





Terrestrial network malfunction

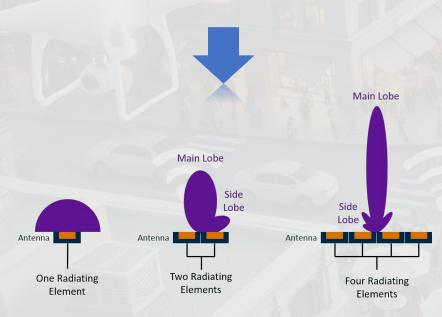








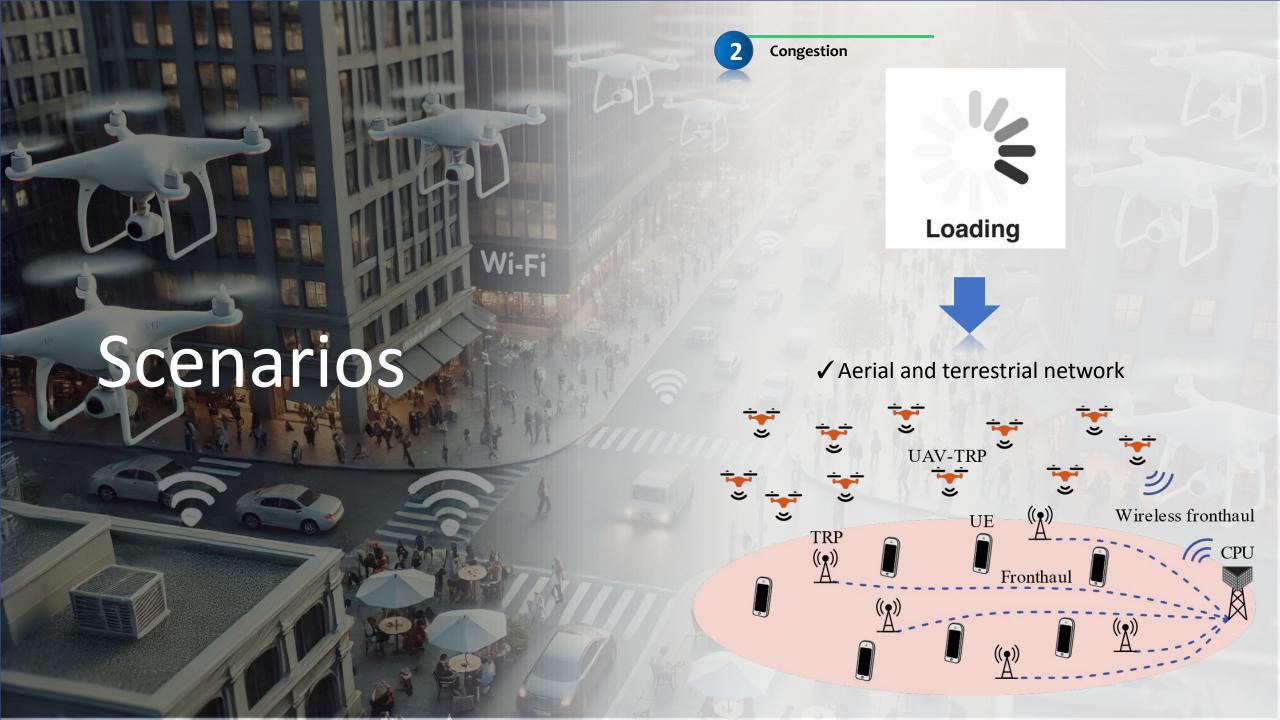
Cooperative UAV swarm



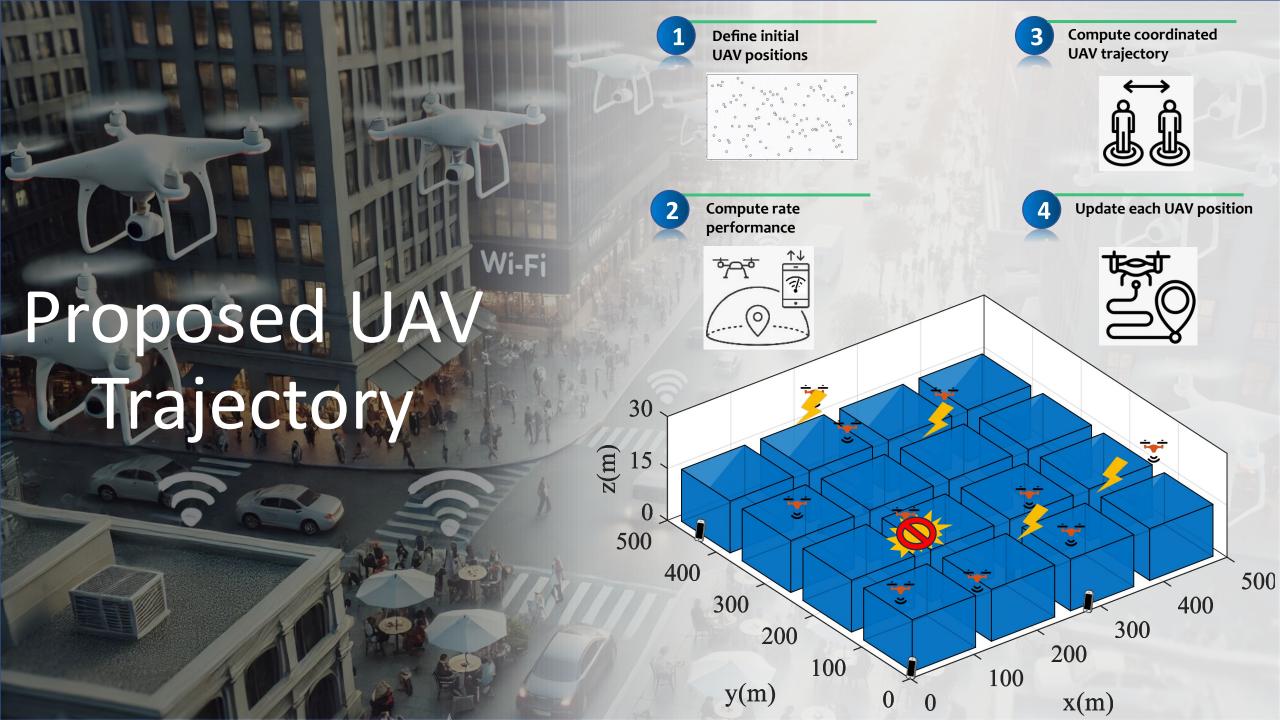
Leads to array gain in D-mMIMO

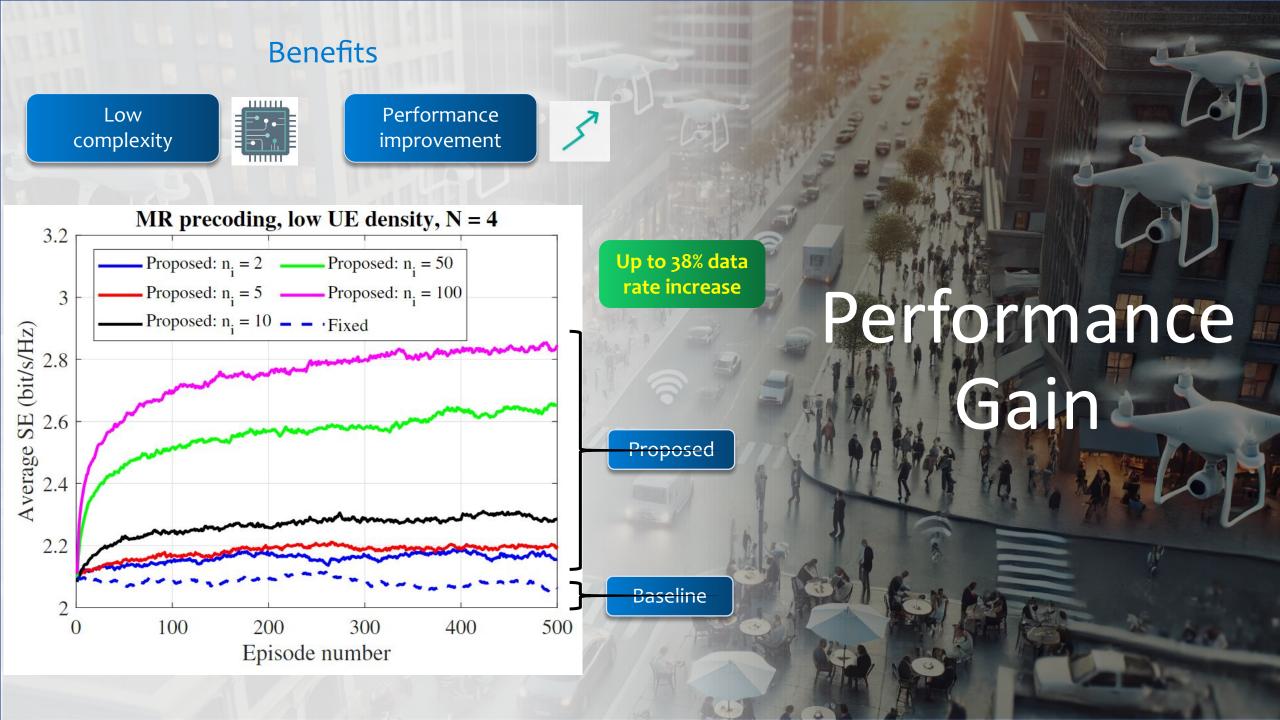


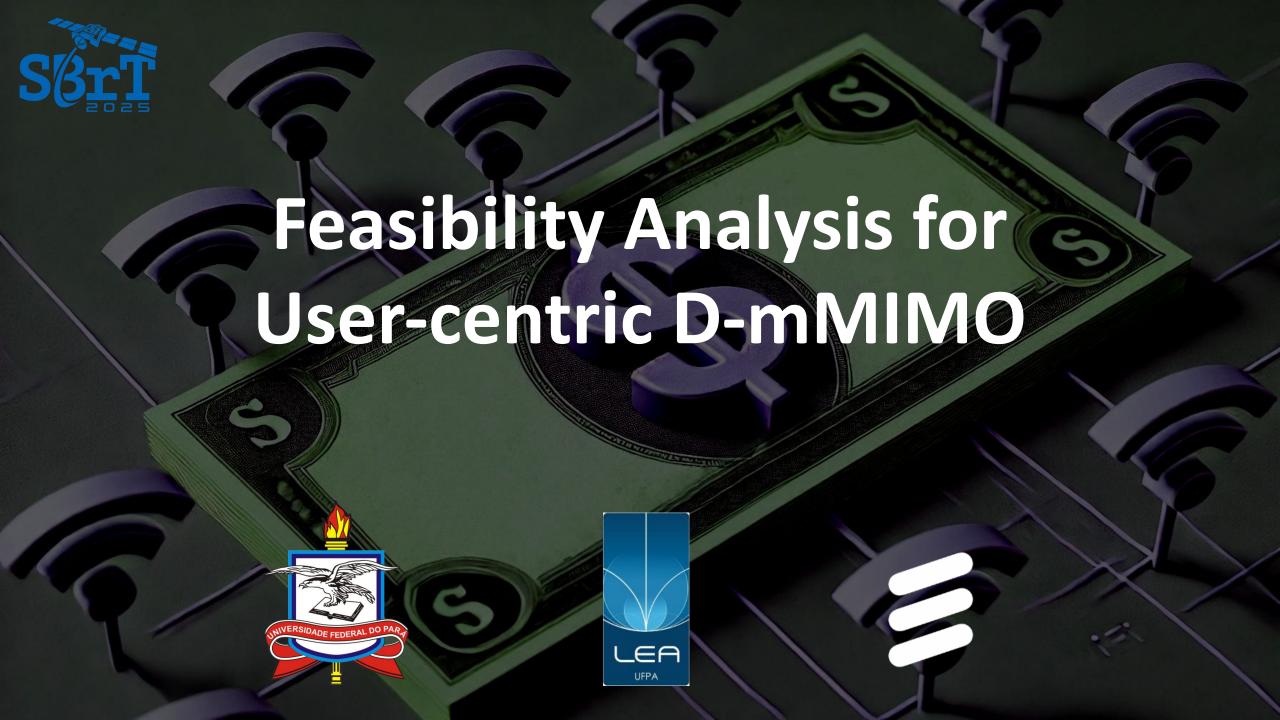


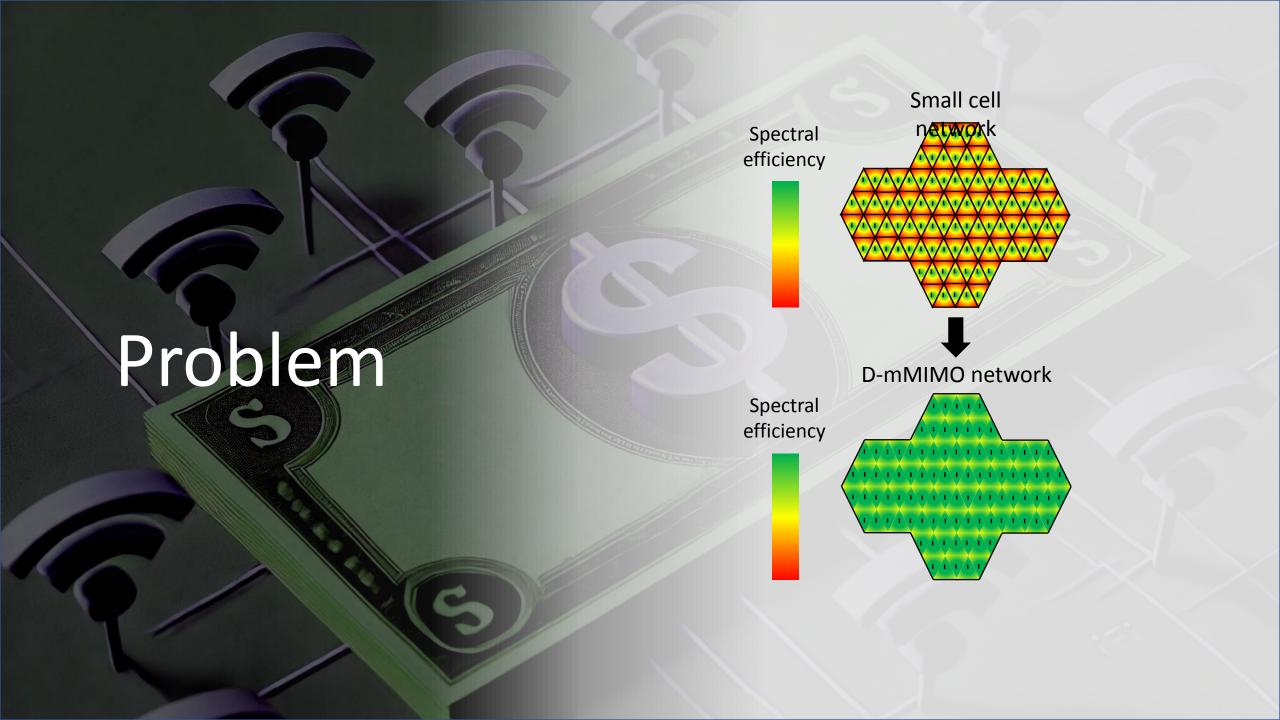


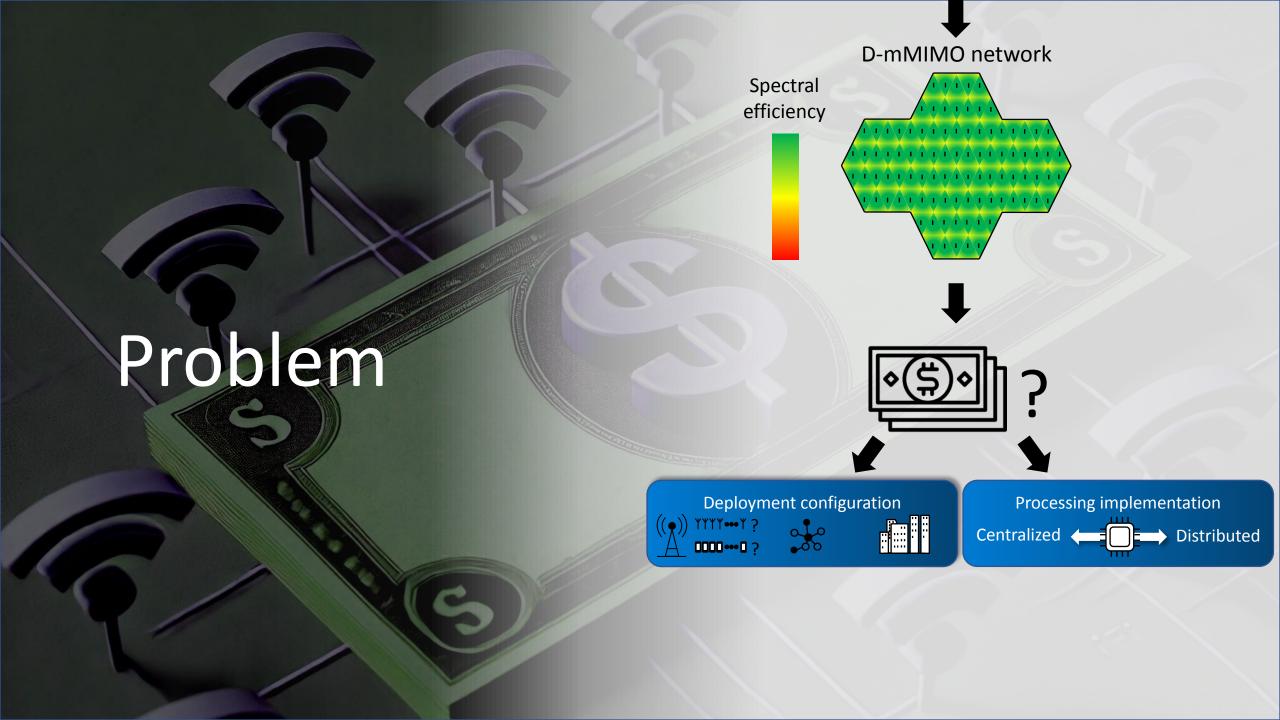


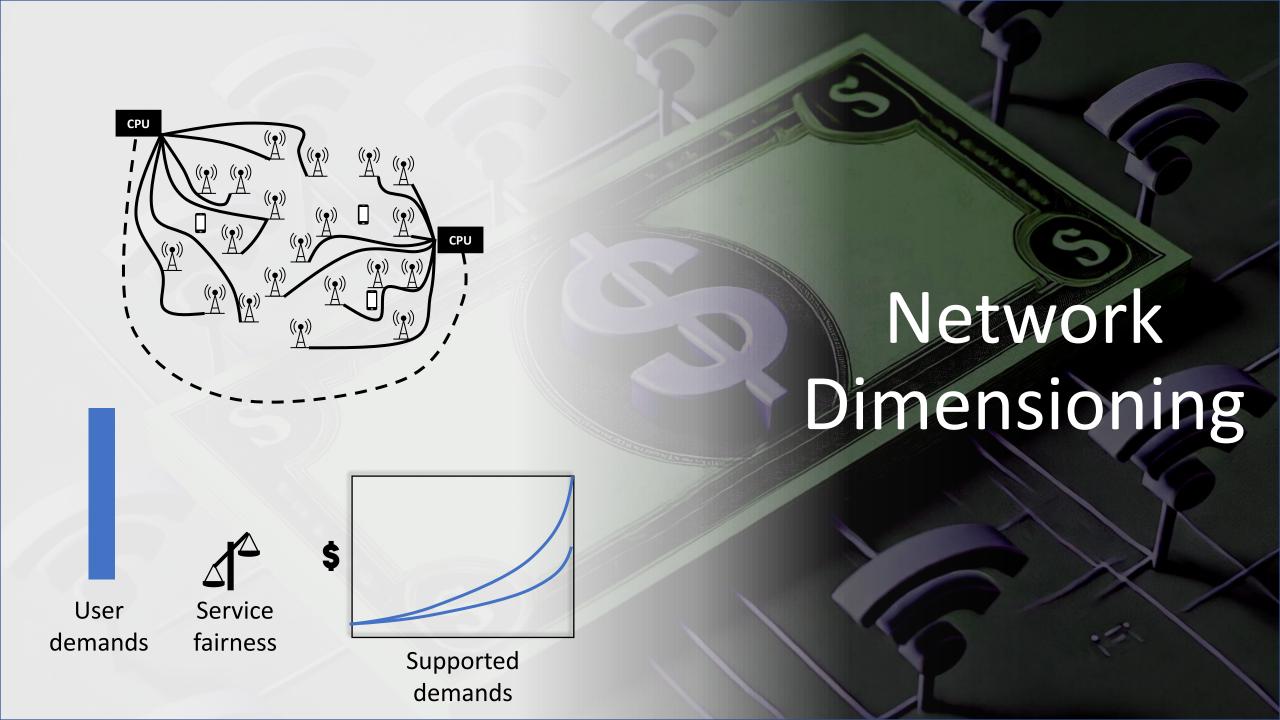


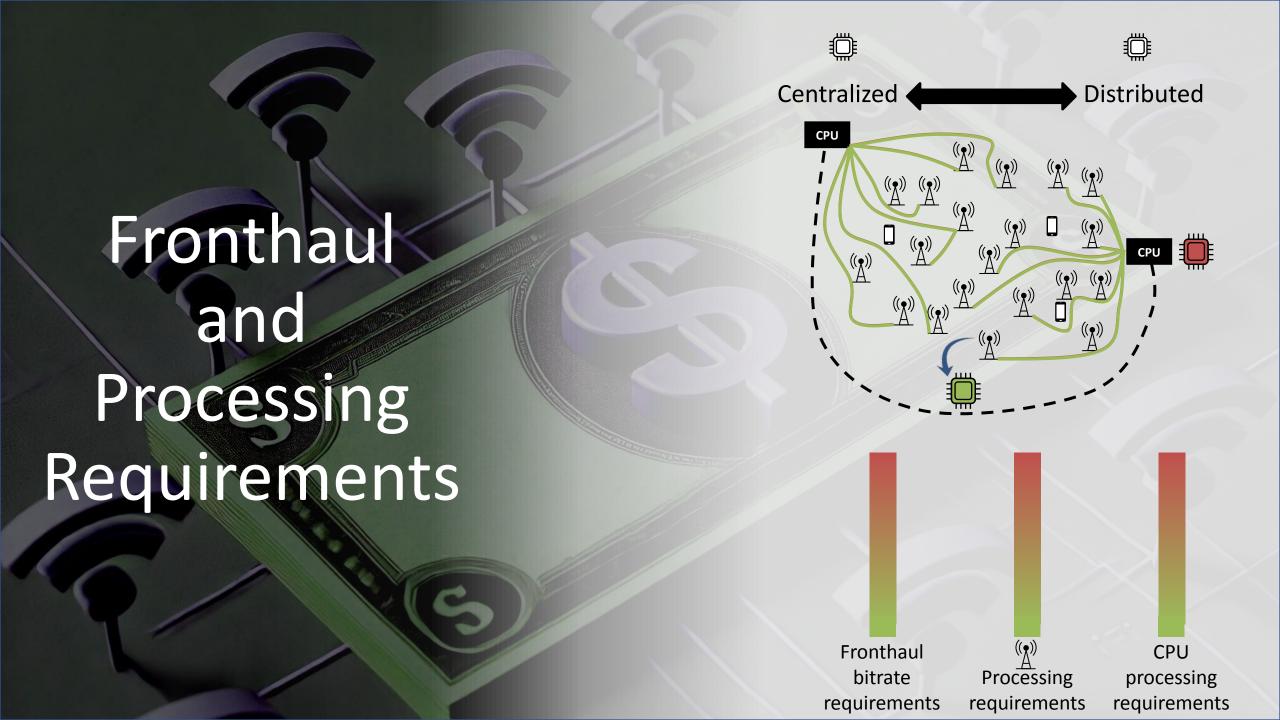


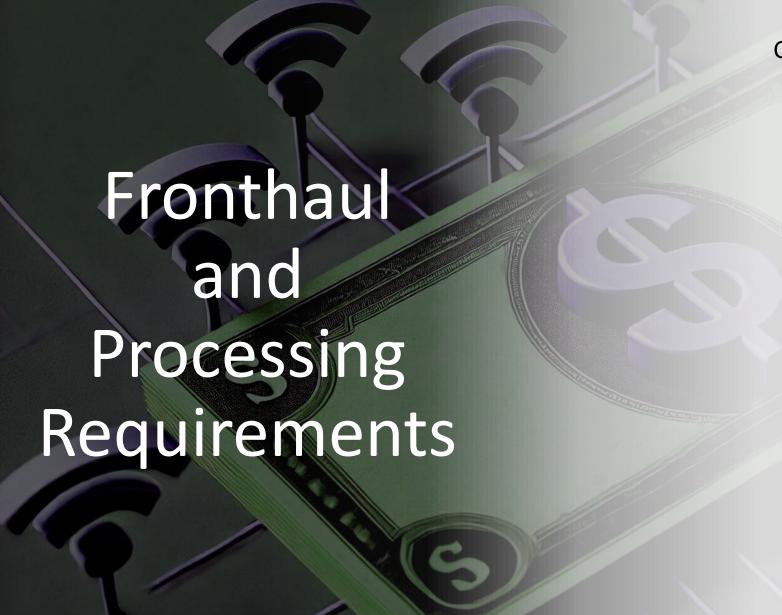


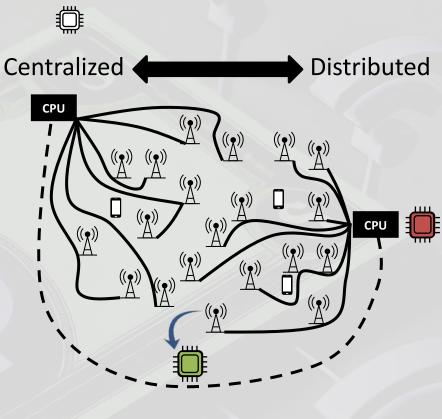










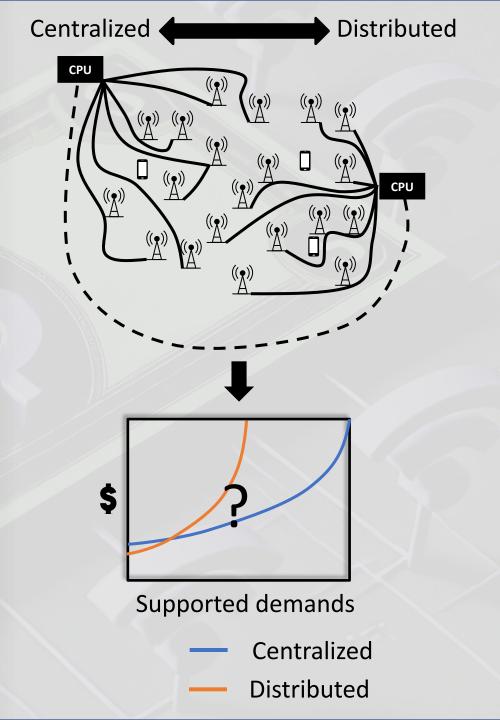


Fronthaul bitrate requirements

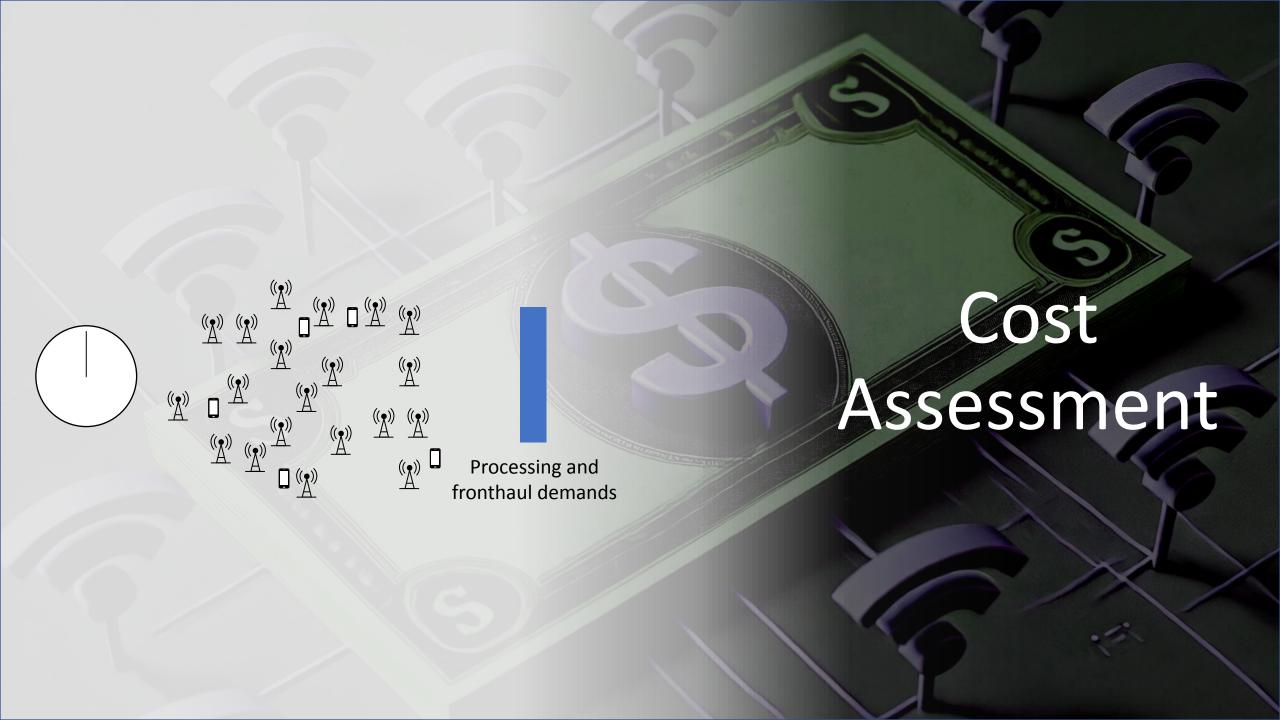
Processing requirements

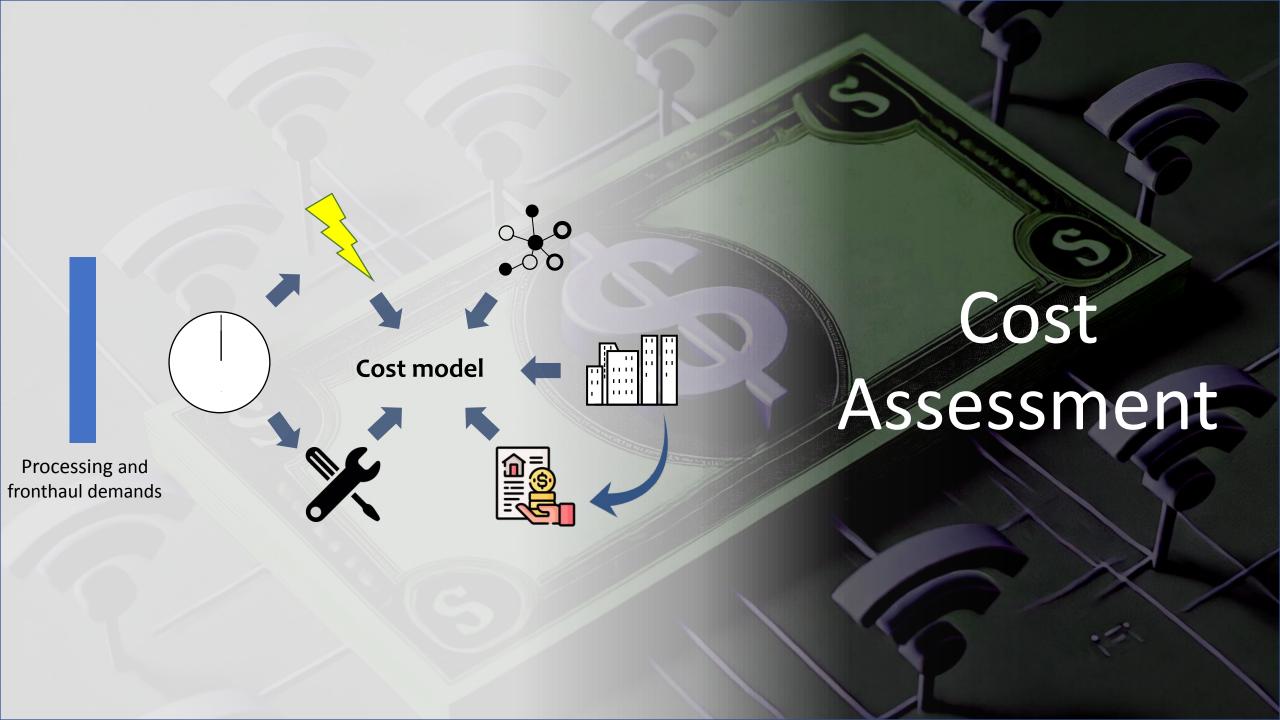
CPU processing requirements

Fronthaul and Processing Requirements









Profitability Distributed vs. Centralized

